



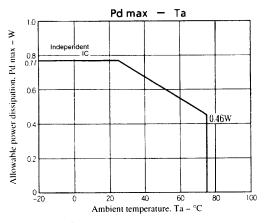
3-Phase Brushless Motor Driver for VCR Capstan Motors

Overview

The LB1980 is a 3-phase brushless motor driver IC that is particularly well suited for driving VCR capstan motors.

Functions

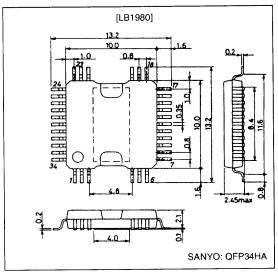
- · Three-phase full wave current linear drive
- Torque ripple correction circuit with variable correction ratio
- · Current limiter circuit
- Output stage up/down oversaturation protection circuit (no external capacitor required)
- · FG amplifier
- · Thermal shutdown circuit



Package Dimensions

unit: mm

3206A-QFP34HA



Specifications

Absolute Maximum Ratings at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
	V _{CC} max		7	V
Maximum supply voltage	V _S max		24	٧
Maximum output current	I _O max		1.3	Α
	Pd max	With an arbitrarily large heat sink	12.5	W
Allowable power dissipation		Independent IC	0.77	W
Operating temperature	Topr		-20 to +75	°C
Storage temperature	Tstg		-55 to +150	ů

Allowable Operating Banges at $Ta = 25^{\circ}C$

Parameter	Symbol Conditions		Ratings	Unit
	Vs		5 to 22	V
Supply voltage	Vcc		4.5 to 5.5	V
Hall input amplitude	VHALL	Between Hall inputs	±30 to ±80	mVo-p
GSENSE input range	V _{GSENSE}	With respect to the control system ground	-0.20 to +0.20	V

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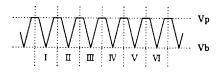


Electrical Characteristics at Ta = 25°C, V_{CC} = 5 $V,\,V_{S}$ = 15 V

Parameter	Symbol	Conditions	Ratings			Unit
	<u> </u>	0000000	min	typ	max	
V _{CC} current drain	lcc	$R_L = \infty$. $V_{CTL} = 0 V$, $V_{LIM} = 0 V$ (When stopped)	-	12	18	mA
[Output]					<u></u>	
Output saturation voltage	Vo(sat)1	I_O = 500 mA, Rf = 0.5 Ω , Sink + Source V_{CTL} = V_{LIM} = 5 V(With saturation protection)	_	2.1	2.6	V
- Corput data and in voltage	Vo(sat)2	I_O = 1.0 A, Rf = 0.5 Ω , Sink + Source V_{CTL} = V_{LIM} = 5 V(With saturation protection)	_	2.6	3.5	v
Output leakage current	lo leak				1.0	mA
[FR]				·	J	
FR pin input threshold voltage	V _{FSR}		2.25	2.50	2.75	Ιv
FR pin input bias current	Ib(FSR)		-5.0	_		μА
[Control]					1	1
CTLREF pin voltage	VC _{REF}		2.05	2.15	2.25	V
CTLREF pin input range	VCREFIN		1.50		3.50	l v
CTL pin input bias current	Ib(CTL)	V _{CTL} = 5 V, with CTLREF open		_	4.0	μА
CTL control start voltage	VCTL(ST)	With Rf = 0.5 Ω , V _{LIM} = 5 V, I _O \geq 10 mA With the Hall input logic fixed(u, v, w = H, H, L)	2.00	2.15	2.30	V
CTL pin control Gm	Gm(CTL)	With Rf = 0.5 Ω , ΔI_O = 200 mA With the Hall input logic fixed(u, v, w = H, H, L)	0.46	0.58	0.70	A/V
[Current Limiter]						
LIM current limiter offset voltage	Voff(LIM)	With Rf = 0.5 Ω , V _{CTL} = 5 V, I _O \geq 10 mA With the Hall input logic fixed(u, v, w = H, H, L)	140	200	260	mV
LIM pin input bias current	lb(LIM)	V _{CTL} = 5 V. CTLREF: OPEN, V _{LIM} = 0 V	-2.5	_	_	μА
LIM pin current limit level	llim	With Rf = 0.5 Ω , V _{CTL} = 5 V, V _{LIM} = 20.6 V With the Hall input logic fixed(u, v, w = H, H, L)	830	900	970	mA
[Hall Amplifier]					L	L
Hall amplifier input offset voltage	Voff(HALL)		-6	_	+6	mV
Hall amplifier input bias current	lb(HALL)		-	1.0	3.0	μА
Hall amplifier common-mode input voltage	Vcm(HALL)		1.3		3.3	V
[TRC]		1.				<u> </u>
Torque ripple correction ratio	TRC	For the troughs and peaks in the Rf waveform at $I_{\rm O}$ = 200 mA(Rf = 0.5 Ω , ADJ: open)(note 1)	-	9	-	%
ADJ pin voltage	Vadj		2.37	2.50	2.63	V
[FG Amplifier]						L
FG amplifier input offset voltage	Voff(FG)		-8	_	+8	mV
FG amplifier input bias current	lb(FG)		-100	_	_	nA
FG amplifier output saturation voltage	Vo(sat)(FG)	For the sink side and the internal pull-up resistor load	-	-	0.5	٧
FG amplifier voltage gain	VG(FG)	For open loop at f = 10 kHz	41.5	44.5	47.5	dB
FG amplifier common-mode input voltage	VCM(FG)		0.5	-	4.0	V
[Saturation]						
Saturation prevention circuit lower side voltage setting	Vo(sat)(DET)	I_O = 10 mA, Rf = 0.5 Ω The voltage between each OUT and Rf when V_{CTL} = V_{LIM} = 5 V.	0.175	0.25	0.325	٧
[TSD]	•	·				
TSD operating temperature	T-TSD	(Design target value)*		180	_ [°C
TSD temperature hysteresis	ΔTSD	(Design target value)*		20		°C
described to the second of the		<u> </u>				

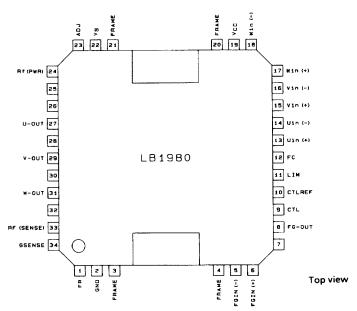
Notes: *Items marked with an asterisk are design target values and are not measured.

1. The torque ripple correction ratio is determined from the Rf voltage waveform as follows.



$$= \frac{2 \cdot (Vp - Vb)}{Vp + Vb} \cdot 100 \ (\%)$$

Pin Assignment



Note: For ground potential stabilization, the FRAME pins must be connected to ground.

Truth Table and Control Functions

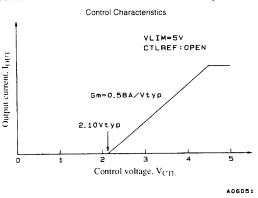
		·			
	Source → Sink	Hall input			FR
	Course / Cirix	U	V	w	
1	$V \rightarrow W$	Н	н	L	H
'	$W \rightarrow V$				Ĺ
2	$U \rightarrow W$		H L		Н
2	W → U	П		_	L
	$U \rightarrow V$	н	L	н	н
3	$V \rightarrow U$			_	L
	$W \rightarrow V$	L	L	н	Н
4	$V \rightarrow W$				L
	W → U	L	н		Н
5 -	$U \rightarrow W$			Н	L
6	$V \rightarrow U$		н		н
	$U \rightarrow V$	L		-	L

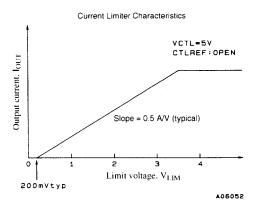
Note: An H in the FR column indicates a voltage of 2.75 V or higher, and an L indicates a voltage of 2.25 V or lower. (When $V_{CC} = 5 \text{ V.}$)

Note: For the Hall inputs. H indicates that the + input is 0.01 V or more higher than the – input, and L indicates that the + input is 0.01 V or more lower than the - input, for all phases.

Note: Since this drive technique is a 180° conducting technique, the phases other than the sink and source phases are never off.

[Control Function and Current Limiter Function]





Pin Functions

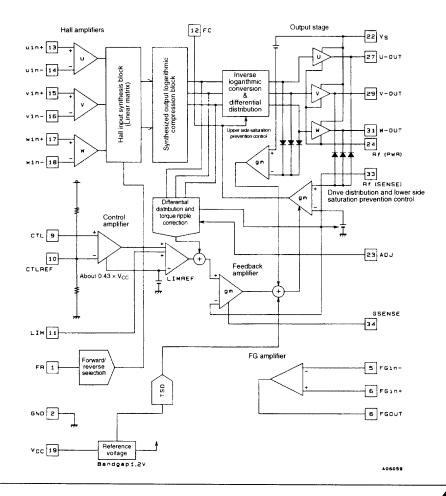
Pin No.	Pin name	Pin functions	Equivalent circuit
1	FR	Forward/reverse selection pin.	
		Forward or reverse is selected by the voltage applied to this pin.	
		(Vth = 2.5 V (typical) at $V_{CC} = 5 \text{ V}$)	Vcc 4
23	ADJ	This pin allows the torque ripple correction ratio to be adjusted	VCC VCC VCC VCC
		externally. Applications that adjust the correction ratio should	Q200#A
		apply a voltage to the ADJ pin with a low-impedance circuit.	FR
		The correction ratio falls as the applied voltage increases, and	2000 FKQ 500Q FKQ
		increases as the applied voltage decreases. The correction ratio	* 71/2 \$10k0 10k0\$
		can be varied over a range from about 0 to twice the ratio when	"
		this pin is left open. (This voltage is set internally to about $V_{\rm CC}/2$,	-
		and the input impedance is about 5 $k\Omega$.)	***************************************
2	GND	Ground for circuits other than the output transistors	
		The output transistor minimum potential is that of the resistor Rf.	
5	FGin-	When the FG amplifier is used as an inverting input amplifier,	W
		a feedback resistor must be inserted between FG-OUT	Vcc A
		and this pin.	
			5 # A Q
			Y
			FRIT ()
6	FGin+	This is the noninverting input when the FG amplifier is used as	FGin (+)
		a differential input amplifier. There is no internally applied bias.	3000
			A05057
8	FG-OUT	This is the FG amplifier output. It has an internal load resistance.	
			Vcc 4 4Vcc 4Vcc
			\$ \$2×0
10			FGOUT ₹10kg
12	FC	Frequency characteristics correction for the speed control loop	3000
l			#
			# # # #
			A0605
9	CTL	Speed control connection. Speed control is implemented as a	
	CIL	Speed control connection. Speed control is implemented as a	
		fixed current drive in which current feedback is applied from Rf. $Gm = 0.58$ A/V (typical) at Rf = 0.5 Ω	
10	CTLREF	Control reference. Although this pin is set to about 0.43 * V _{CC}	Acc Acc Acc
'	JILILI	internally, it can be changed by applying a voltage with	CTL \$10k0 200MA
		a low-impedance circuit. (It has an input impedance of about	2000 CTLREF MAX
		4.3 k Ω (typical).)	7.5ku = 1 LIM
		· (-) plos()/)	" Joona " J 2000
11	LIM	Current limiter function control. The voltage applied to this pin	
	- ***	changes the output current linearly. Slope = 0.5 A/V (typical)	A06035
		at Rf = 0.5 Ω .	
13	Uin+	U phase Hall element input. IN+ must be higher than IN-	Fact (1) fact
14	Uin-	for a logical high input level.	Each (+) input Each (-) input
15	Vin+	V phase Hall element input. IN+ must be higher than IN-	2000
16	Vin-	for a logical high input level.	*
17	Win+	W phase Hall element input. IN+ must be higher than IN-	# S100#A #
18	Win-	for a logical high input level.	,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,
		J J	/// A05053

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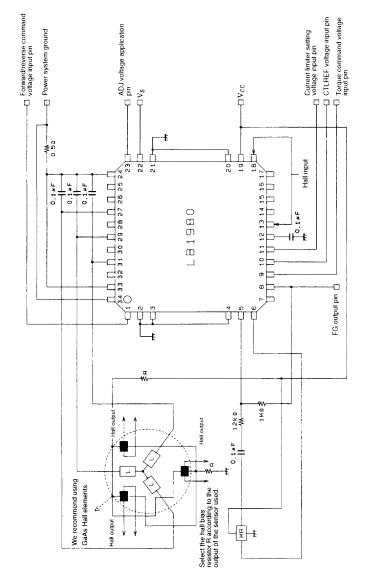
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Pin No.	Pin name	Pin functions	Equivalent circuit
19	V _{CC}	Power supply for all IC internal circuits except for the cutput	
		block. This voltage must be stabilized so that ripple and noise	
		do not enter the IC circuits.	
22	Vs	Output block power supply	
24	Rf	These are the output current detection pins, and the control	
	(PWR)	block current feedback is applied by connecting the resistor	
33	Rf	Rf between these pins and ground. Additionally, the lower side	Vs Vcc
	(SENSE)	oversaturation protection circuit and the torque ripple correction	Lower side saturation
		circuit operate according to the voltages on these pins.	prevention circuit input block
		In particular, since the oversaturation protection level is set by	AVCC
		the voltages on these pins, if the value of Rf is lowered	2000
		excessively, the effectiveness of the lower side oversaturation	→ \$30k0 # \$10#A
		protection in the high-current region will be degraded.	2000
		Note that the POWER and SENSE pins must be connected	Af (SENSE)
		together.	Rf (POWER)
27	U-OUT	U phase output. (A spark killer diode is built in.)	A08054
29	V-OUT	V phase output. (A spark killer diode is built in.)	ļ
31	W-OUT	W phase output. (A spark killer diode is built in.)	
34	GSENSE	Ground sense. By connecting this pin near the ground for the	
		Rf resistor side of the motor ground wiring that includes Rf,	
		the influence that the common ground impedance has on	
		Rf can be removed. (This pin must not be left open.)	

Block Diagram



Sample Application Circuit



Note: The constants presented in this sample application circuit are examples, and the circuit characteristics for these values are not guaranteed.