

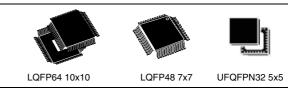
STM32F051x4 STM32F051x6 STM32F051x8

Low- and medium-density advanced ARM[™]-based 32-bit MCU with 16 to 64 Kbytes Flash, timers, ADC, DAC and comm. interfaces

Datasheet - production data

Features

- Operating conditions:
 - Voltage range: 2.0 V to 3.6 V
- ARM 32-bit Cortex®-M0 CPU (48 MHz max)
- Memories
 - 16 to 64 Kbytes of Flash memory
 - 8 Kbytes of SRAM with HW parity checking
- CRC calculation unit
- Clock management
 - 4 to 32 MHz crystal oscillator
 - 32 kHz oscillator for RTC with calibration
 - Internal 8 MHz RC with x6 PLL option
 - Internal 40 kHz RC oscillator
- Calendar RTC with alarm and periodic wakeup from Stop/Standby
- Reset and supply management
 - Power-on/Power down reset (POR/PDR)
 - Programmable voltage detector (PVD)
- Low power Sleep, Stop, and Standby modes
- V_{BAT} supply for RTC and backup registers
- 5-channel DMA controller
- 1×12 -bit, 1.0 µs ADC (up to 16 channels)
 - Conversion range: 0 to 3.6V
 - Separate analog supply from 2.4 up to 3.6
- Two fast low-power analog comparators with programmable input and output
- One 12-bit D/A converter
- Up to 55 fast I/Os
 - All mappable on external interrupt vectors
 - Up to 36 I/Os with 5 V tolerant capability
- Up to 18 capacitive sensing channels supporting touchkey, linear and rotary touch sensors
- 96-bit unique ID
- Serial wire debug (SWD)



■ Up to 11 timers

- One 16-bit 7-channel advanced-control timer for 6 channels PWM output, with deadtime generation and emergency stop
- One 32-bit and one 16-bit timer, with up to 4 IC/OC, usable for IR control decoding
- One 16-bit timer, with 2 IC/OC, 1 OCN, deadtime generation and emergency stop
- Two 16-bit timers, each with IC/OC and OCN, deadtime generation, emergency stop and modulator gate for IR control
- One 16-bit timer with 1 IC/OC
- Independent and system watchdog timers
- SysTick timer: 24-bit downcounter
- One 16-bit basic timer to drive the DAC

■ Communication interfaces

- Up to two I²C interfaces; one supporting Fast Mode Plus (1 Mbit/s) with 20 mA current sink, SMBus/PMBus, and wakeup from STOP
- Up to two USARTs supporting master synchronous SPI and modem control; one with ISO7816 interface, LIN, IrDA capability, auto baud rate detection and wakeup feature
- Up to two SPIs (18 Mbit/s) with 4 to 16 programmable bit frame, 1 with I²S interface multiplexed
- HDMI CEC interface, wakeup on header reception

Table 1. Device summary

Reference	Part number
STM32F051x4	STM32F051K4, STM32F051C4, STM32F051R4
STM32F051x6	STM32F051K6, STM32F051C6, STM32F051R6
STM32F051x8	STM32F051C8, STM32F051R8, STM32F051K8

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Introduction STM32F051x

1 Introduction

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This datasheet provides the ordering information and mechanical device characteristics of the STM32F051xx microcontrollers.

This STM32F051x6 and STM32F051x8 datasheet should be read in conjunction with the STM32F051xx reference manual. The reference manual is available from the STMicroelectronics website www.st.com.

For information on the ARM Cortex[™]-M0 core please refer to the Cortex[™]-M0 Technical Reference Manual, available from the www.arm.com website at the following address: http://infocenter.arm.com/help/index.jsp?topic=/com.arm.doc.ddi0432c/index.html.



STM32F051x Description

2 Description

The STM32F051xx family incorporates the high-performance ARM Cortex™-M0 32-bit RISC core operating at a 48 MHz frequency, high-speed embedded memories (Flash memory up to 64 Kbytes and SRAM up to 8 Kbytes), and an extensive range of enhanced peripherals and I/Os. All devices offer standard communication interfaces (up to two I²Cs, two SPIs, one I2S, one HDMI CEC, and up to two USARTs), one 12-bit ADC, one 12-bit DAC, up to five general-purpose 16-bit timers, a 32-bit timer and an advanced-control PWM timer.

The STM32F051xx family operates in the -40 to +85 °C and -40 to +105 °C temperature ranges, from a 2.0 to 3.6 V power supply. A comprehensive set of power-saving modes allows the design of low-power applications.

The STM32F051xx family includes devices in three different packages ranging from 32 pins to 64 pins. Depending on the device chosen, different sets of peripherals are included. The description below provides an overview of the complete range of peripherals proposed in this family.

These features make the STM32F051xx microcontroller family suitable for a wide range of applications such as application control and user interfaces, handheld equipment, A/V receivers and digital TV, PC peripherals, gaming and GPS platforms, industrial applications, PLCs, inverters, printers, scanners, alarm systems, video intercoms, and HVACs.

Description STM32F051x

Table 2. STM32F051xx family device features and peripheral counts

Peripheral		STM32F051Kx			ST	STM32F051Cx			STM32F051Rx		
Flash (Kbyt	es)	16	32	64	16	32	64	16	32	64	
SRAM (Kby	rtes)	4	4 8 4 8					4	4	8	
	Advanced control	1 (16-bit)									
Timers	General purpose		5 (16-bit) 1 (32-bit)								
	Basic					1 (16-bit)					
	SPI [I2S] ⁽¹⁾		1[1] ⁽²⁾		1[1]] (2)	2[1]	1[1] (2)	2[1]	
Comm.	I ² C		1 ⁽³⁾		1	(3)	2	1	(3)	2	
interfaces	USART	1 ⁽⁴⁾	2	2	1 ⁽⁴⁾ 2		2	1 ⁽⁴⁾		2	
	CEC				1	1		•			
12-bit synchronized ADC (number of channels)		1 1 (10 ext. + 3 int.) (16 ext. + 3 in						int.)			
GPIOs			27 39				55				
Capacitive channels	sensing	14 17					18				
12-bit DAC (number of	channels)	1 (1)									
Analog con	nparator	2									
Max. CPU	frequency	48 MHz									
Operating v	oltage	2.0 to 3.6 V									
Operating temperature		Ambient operating temperature: -40 °C to 85 °C / -40 °C to 105 °C Junction temperature: -40 °C to 125 °C									
Packages		ι	JFQFPN3	2		LQFP48			LQFP64		

^{1.} The SPI1 interface can be used either in SPI mode or in I2S audio mode.

^{2.} SPI2 is not present

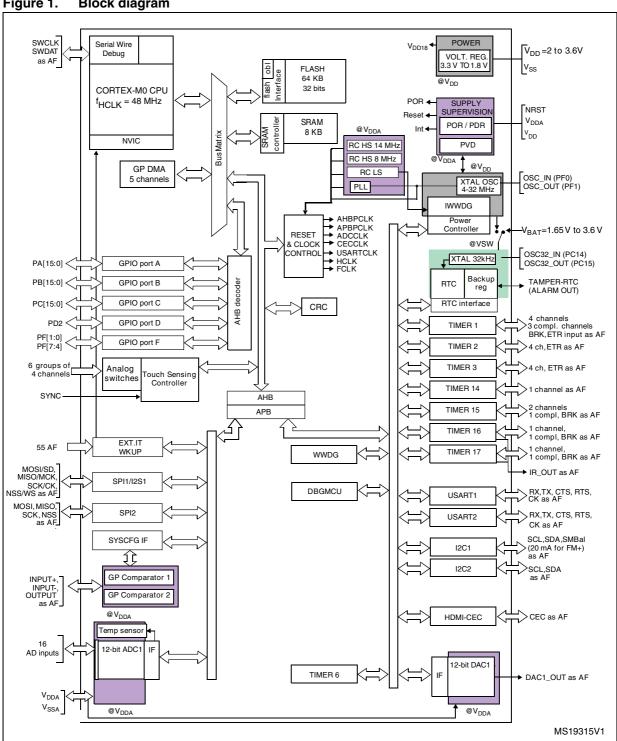
^{3.} I2C2 is not present

^{4.} USART2 is not present

STM32F051x **Device overview**

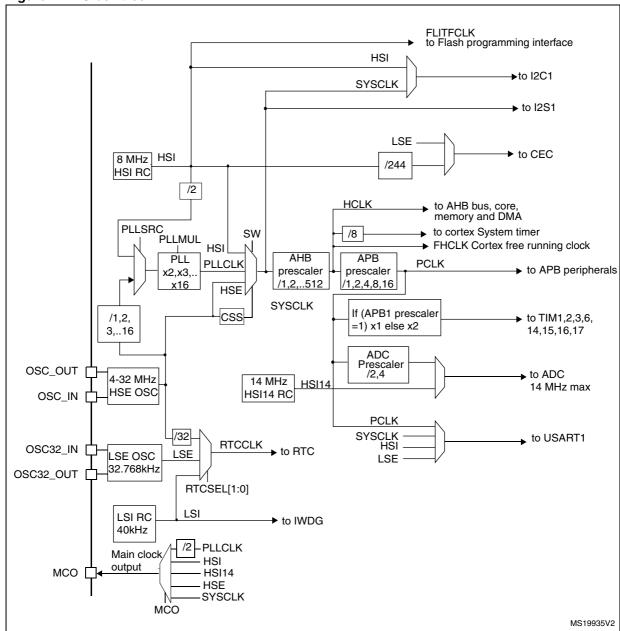
3 **Device overview**

Figure 1. **Block diagram**



Device overview STM32F051x

Figure 2. Clock tree



4 Functional overview

4.1 ARM® CortexTM-M0 core with embedded Flash and SRAM

The ARM CortexTM-M0 processor is the latest generation of ARM processors for embedded systems. It has been developed to provide a low-cost platform that meets the needs of MCU implementation, with a reduced pin count and low-power consumption, while delivering outstanding computational performance and an advanced system response to interrupts.

The ARM Cortex[™]-M0 32-bit RISC processor features exceptional code-efficiency, delivering the high-performance expected from an ARM core in the memory size usually associated with 8- and 16-bit devices.

The STM32F051xx family has an embedded ARM core and is therefore compatible with all ARM tools and software.

Figure 1 shows the general block diagram of the device family.

4.2 Memories

The device has the following features:

- Up to 8 Kbytes of embedded SRAM accessed (read/write) at CPU clock speed with 0 wait states and featuring embedded parity checking with exception generation for failcritical applications.
- The non-volatile memory is divided into two arrays:
 - 16 to 64 Kbytes of embedded Flash memory for programs and data
 - Option bytes

The option bytes are used to write-protect the memory (with 4 KB granularity) and/or readout-protect the whole memory with the following options:

- Level 0: no readout protection
- Level 1: memory readout protection, the Flash memory cannot be read from or written to if either debug features are connected or boot in RAM is selected
- Level 2: chip readout protection, debug features (Cortex-M0 serial wire) and boot in RAM selection disabled

4.3 Cyclic redundancy check calculation unit (CRC)

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code from a 96-bit data word and a fixed generator polynomial.

Among other applications, CRC-based techniques are used to verify data transmission or storage integrity. In the scope of the EN/IEC 60335-1 standard, they offer a means of verifying the Flash memory integrity. The CRC calculation unit helps compute a signature of the software during runtime, to be compared with a reference signature generated at link-time and stored at a given memory location.

4.4 Direct memory access controller (DMA)

The 5-channel general-purpose DMAs manage memory-to-memory, peripheral-to-memory and memory-to-peripheral transfers.

The DMA supports circular buffer management, removing the need for user code intervention when the controller reaches the end of the buffer.

Each channel is connected to dedicated hardware DMA requests, with support for software trigger on each channel. Configuration is made by software and transfer sizes between source and destination are independent.

DMA can be used with the main peripherals: SPI, I2S, I2C, USART, all TIMx timers (except TIM14), DAC and ADC.

4.5 Nested vectored interrupt controller (NVIC)

The STM32F051xx family embeds a nested vectored interrupt controller able to handle up to 32 maskable interrupt channels (not including the 16 interrupt lines of Cortex[™]-M0) and 16 priority levels.

- Closely coupled NVIC gives low latency interrupt processing
- Interrupt entry vector table address passed directly to the core
- Closely coupled NVIC core interface
- Allows early processing of interrupts
- Processing of late arriving higher priority interrupts
- Support for tail-chaining
- Processor state automatically saved
- Interrupt entry restored on interrupt exit with no instruction overhead

This hardware block provides flexible interrupt management features with minimal interrupt latency.

4.6 Extended interrupt/event controller (EXTI)

The external interrupt/event controller consists of 24 edge detector lines used to generate interrupt/event requests and wake-up the system. Each line can be independently configured to select the trigger event (rising edge, falling edge, both) and can be masked independently. A pending register maintains the status of the interrupt requests. The EXTI can detect an external line with a pulse width shorter than the internal clock period. Up to 55 GPIOs can be connected to the 16 external interrupt lines.

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4.7 Clocks and startup

System clock selection is performed on startup, however the internal RC 8 MHz oscillator is selected as default CPU clock on reset. An external 4-32 MHz clock can be selected, in which case it is monitored for failure. If failure is detected, the system automatically switches back to the internal RC oscillator. A software interrupt is generated if enabled. Similarly, full interrupt management of the PLL clock entry is available when necessary (for example on failure of an indirectly used external crystal, resonator or oscillator).

Several prescalers allow the application to configure the frequency of the AHB and the APB domains. The maximum frequency of the AHB and the APB domains is 48 MHz.

4.8 Boot modes

At startup, the boot pin and boot selector option bit are used to select one of three boot options:

- Boot from User Flash
- Boot from System Memory
- Boot from embedded SRAM

The boot loader is located in System Memory. It is used to reprogram the Flash memory by using USART1.

4.9 Power management

4.9.1 Power supply schemes

- $V_{DD} = 2.0$ to 3.6 V: external power supply for I/Os and the internal regulator. Provided externally through V_{DD} pins.
- V_{DDA} = 2.0 to 3.6 V: external analog power supply for ADC, Reset blocks, RCs and PLL (minimum voltage to be applied to V_{DDA} is 2.4 V when the ADC and DAC are used).
 The V_{DDA} voltage level must be always greater or equal to the V_{DD} voltage level and must be provided first.
- V_{BAT} = 1.65 to 3.6 V: power supply for RTC, external clock 32 kHz oscillator and backup registers (through power switch) when V_{DD} is not present.

For more details on how to connect power pins, refer to Figure 9: Power supply scheme.

4.9.2 Power supply supervisors

The device has integrated power-on reset (POR) and power-down reset (PDR) circuits. They are always active, and ensure proper operation above a threshold of 2 V. The device remains in reset mode when the monitored supply voltage is below a specified threshold, $V_{POR/PDR}$, without the need for an external reset circuit.

- The POR monitors only the V_{DD} supply voltage. During the startup phase it is required that V_{DDA} should arrive first and be greater than or equal to V_{DD}.
- The PDR monitors both the V_{DD} and V_{DDA} supply voltages, however the V_{DDA} power supply supervisor can be disabled (by programming a dedicated Option bit) to reduce the power consumption if the application design ensures that V_{DDA} is higher than or equal to V_{DD}.

The device features an embedded programmable voltage detector (PVD) that monitors the V_{DD} power supply and compares it to the V_{PVD} threshold. An interrupt can be generated when V_{DD} drops below the V_{PVD} threshold and/or when V_{DD} is higher than the V_{PVD} threshold. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state. The PVD is enabled by software.

4.9.3 Voltage regulator

The regulator has three operating modes: main (MR), low power (LPR) and power down.

- MR is used in normal operating mode (Run)
- LPR can be used in Stop mode where the power demand is reduced
- Power down is used in Standby mode: the regulator output is in high impedance: the kernel circuitry is powered down, inducing zero consumption (but the contents of the registers and SRAM are lost)

This regulator is always enabled after reset. It is disabled in Standby mode, providing high impedance output.

4.10 Low-power modes

The STM32F051xx family supports three low-power modes to achieve the best compromise between low power consumption, short startup time and available wakeup sources:

Sleep mode

In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs.

Stop mode

Stop mode achieves very low power consumption while retaining the content of SRAM and registers. All clocks in the 1.8 V domain are stopped, the PLL, the HSI RC and the HSE crystal oscillators are disabled. The voltage regulator can also be put either in normal or in low power mode.

The device can be woken up from Stop mode by any of the EXTI lines. The EXTI line source can be one of the 16 external lines, the PVD output, RTC alarm, COMPx, I2C1, USART1 or the CEC.

The I2C1, USART1 and the CEC can be configured to enable the HSI RC oscillator for processing incoming data. If this is used, the voltage regulator should not be put in the low-power mode but kept in normal mode.

Standby mode

The Standby mode is used to achieve the lowest power consumption. The internal voltage regulator is switched off so that the entire 1.8 V domain is powered off. The PLL, the HSI RC and the HSE crystal oscillators are also switched off. After entering Standby mode, SRAM and register contents are lost except for registers in the Backup domain and Standby circuitry.

The device exits Standby mode when an external reset (NRST pin), a IWDG reset, a rising edge on the WKUP pins, or an RTC alarm occurs.

Note: The RTC, the IWDG, and the corresponding clock sources are not stopped by entering Stop or Standby mode.

4.11 Real-time clock (RTC) and backup registers

The RTC and the 5 backup registers are supplied through a switch that takes power either on V_{DD} supply when present or through the V_{BAT} pin. The backup registers are five 32-bit registers used to store 20 bytes of user application data when V_{DD} power is not present. They are not reset by a system or power reset, or when the device wakes up from Standby mode.

The RTC is an independent BCD timer/counter. Its main features are the following:

- Calendar with subsecond, seconds, minutes, hours (12 or 24 format), week day, date, month, year, in BCD (binary-coded decimal) format.
- Automatically correction for 28, 29 (leap year), 30, and 31 day of the month.
- Programmable alarm with wake up from Stop and Standby mode capability.
- On-the-fly correction from 1 to 32767 RTC clock pulses. This can be used to synchronize it with a master clock.
- Digital calibration circuit with 1 ppm resolution, to compensate for quartz crystal inaccuracy.
- 2 anti-tamper detection pins with programmable filter. The MCU can be woken up from Stop and Standby modes on tamper event detection.
- Timestamp feature which can be used to save the calendar content. This function can triggered by an event on the timestamp pin, or by a tamper event. The MCU can be woken up from Stop and Standby modes on timestamp event detection.

The RTC clock sources can be:

- A 32.768 kHz external crystal
- A resonator or oscillator
- The internal low-power RC oscillator (typical frequency of 40 kHz)
- The high-speed external clock divided by 32.

4.12 Timers and watchdogs

The STM32F051xx family devices include up to six general-purpose timers, one basic timer and an advanced control timer.

Table 3 compares the features of the advanced-control, general-purpose and basic timers.

Table 3. Timer feature comparison

Timer type	Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/compare channels	Complementary outputs
Advanced control	TIM1	16-bit	Up, down, up/down	Any integer between 1 and 65536	Yes	4	Yes
	TIM2	32-bit	Up, down, up/down	Any integer between 1 and 65536	Yes	4	No
	TIM3	16-bit	Up, down, up/down	Any integer between 1 and 65536	Yes	4	No
General purpose	TIM14	16-bit	Up	Any integer between 1 and 65536	No	1	No
	TIM15	16-bit	Up	Any integer between 1 and 65536	Yes	2	Yes
	TIM16, TIM17	16-bit	Up	Any integer between 1 and 65536	Yes	1	Yes
Basic	TIM6	16-bit	Up	Any integer between 1 and 65536	Yes	0	No

4.12.1 Advanced-control timer (TIM1)

The advanced-control timer (TIM1) can be seen as a three-phase PWM multiplexed on 6 channels. It has complementary PWM outputs with programmable inserted dead times. It can also be seen as a complete general-purpose timer. The 4 independent channels can be used for:

- Input capture
- Output compare
- PWM generation (edge or center-aligned modes)
- One-pulse mode output

If configured as a standard 16-bit timer, it has the same features as the TIMx timer. If configured as the 16-bit PWM generator, it has full modulation capability (0-100%).

The counter can be frozen in debug mode.

Many features are shared with those of the standard timers which have the same architecture. The advanced control timer can therefore work together with the other timers via the Timer Link feature for synchronization or event chaining.

4.12.2 General-purpose timers (TIM2..3, TIM14..17)

There are six synchronizable general-purpose timers embedded in the STM32F051xx devices (see *Table 3* for differences). Each general-purpose timer can be used to generate PWM outputs, or as simple time base.

TIM2, TIM3

STM32F051xx devices feature two synchronizable 4-channel general-purpose timers. TIM2 is based on a 32-bit auto-reload up/downcounter and a 16-bit prescaler. TIM3 is based on a 16-bit auto-reload up/downcounter and a 16-bit prescaler. They feature 4 independent channels each for input capture/output compare, PWM or one-pulse mode output. This gives up to 12 input captures/output compares/PWMs on the largest packages.

The TIM2 and TIM3 general-purpose timers can work together or with the TIM1 advanced-control timer via the Timer Link feature for synchronization or event chaining.

TIM2 and TIM3 both have independent DMA request generation.

These timers are capable of handling quadrature (incremental) encoder signals and the digital outputs from 1 to 3 hall-effect sensors.

Their counters can be frozen in debug mode.

TIM14

This timer is based on a 16-bit auto-reload upcounter and a 16-bit prescaler.

TIM14 features one single channel for input capture/output compare, PWM or one-pulse mode output.

Its counter can be frozen in debug mode.

TIM15, TIM16 and TIM17

These timers are based on a 16-bit auto-reload upcounter and a 16-bit prescaler.

TIM15 has two independent channels, whereas TIM16 and TIM17 feature one single channel for input capture/output compare, PWM or one-pulse mode output.

The TIM15, TIM16 and TIM17 timers can work together, and TIM15 can also operate with TIM1 via the Timer Link feature for synchronization or event chaining.

TIM15 can be synchronized with TIM16 and TIM17.

TIM15, TIM16, and TIM17 have a complementary output with dead-time generation and independent DMA request generation

Their counters can be frozen in debug mode.

4.12.3 Basic timer TIM6

This timer is mainly used for DAC trigger generation. It can also be used as a generic 16-bit time base.

4.12.4 Independent watchdog (IWDG)

The independent watchdog is based on an 8-bit prescaler and 12-bit downcounter with user-defined refresh window. It is clocked from an independent 40 kHz internal RC and as it operates independently from the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free running timer for application timeout management. It is hardware or software configurable through the option bytes. The counter can be frozen in debug mode.

4.12.5 System window watchdog (WWDG)

The system window watchdog is based on a 7-bit downcounter that can be set as free running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from the APB clock (PCLK). It has an early warning interrupt capability and the counter can be frozen in debug mode.

4.12.6 SysTick timer

This timer is dedicated to real-time operating systems, but could also be used as a standard down counter. It features:

- A 24-bit down counter
- Autoreload capability
- Maskable system interrupt generation when the counter reaches 0.
- Programmable clock source (HCLK or HCLK/8)

4.13 Inter-integrated circuit interfaces (I²C)

Up to two I^2C interfaces (I2C1 and I2C2) can operate in multimaster or slave modes. Both can support Standard mode (up to 100 kbit/s) or Fast mode (up to 400 kbit/s) and I2C1 supports also Fast Mode Plus (up to 1 Mbit/s) with 20 mA output drive.

Both support 7-bit and 10-bit addressing modes, multiple 7-bit slave addresses (2 addresses, 1 with configurable mask). They also include programmable analog and digital noise filters.

	Analog filter	Digital filter
Pulse width of suppressed spikes	≥ 50 ns	Programmable length from 1 to 15 I2C peripheral clocks
Benefits	Available in Stop mode	Extra filtering capability vs. standard requirements. Stable length
Drawbacks	Variations depending on temperature, voltage, process	Disabled when Wakeup from Stop mode is enabled

In addition, I2C1 provides hardware support for SMBUS 2.0 and PMBUS 1.1: ARP capability, Host notify protocol, hardware CRC (PEC) generation/verification, timeouts verifications and ALERT protocol management. I2C1 also has a clock domain independent from the CPU clock, allowing the I2C1 to wake up the MCU from Stop mode on address match.

The I2C interfaces can be served by the DMA controller.

Refer to *Table 5* for the differences between I2C1 and I2C2.

Table 5. STM32F051xx I²C implementation

I2C features ⁽¹⁾	I2C1	I2C2
7-bit addressing mode	Х	Х
10-bit addressing mode	Х	Х
Standard mode (up to 100 kbit/s)	Х	Х
Fast mode (up to 400 kbit/s)	Х	Х
Fast Mode Plus with 20mA output drive I/Os (up to 1 Mbit/s)	Х	
Independent clock	Х	
SMBus	Х	
Wakeup from STOP	Х	

^{1.} X = supported.

4.14 Universal synchronous/asynchronous receiver transmitters (USART)

The device embeds up to two universal synchronous/asynchronous receiver transmitters (USART1 and USART2), which communicate at speeds of up to 6 Mbit/s.

They provide hardware management of the CTS, RTS and RS485 DE signals, multiprocessor communication mode, master synchronous communication and single-wire half-duplex communication mode. The USART1 supports also SmartCard communication (ISO 7816), IrDA SIR ENDEC, LIN Master/Slave capability, auto baud rate feature and has a clock domain independent from the CPU clock, allowing the USART1 to wake up the MCU from Stop mode.

The USART interfaces can be served by the DMA controller. Serial peripheral interface (SPI).

Refer to Table 6 for the differences between USART1 and USART2.

Table 6. STM32F051xx USART implementation

USART modes/features ⁽¹⁾	USART1	USART2
Hardware flow control for modem	X	X
Continuous communication using DMA	Х	Х
Multiprocessor communication	Х	Х
Synchronous mode	Х	Х
Smartcard mode	Х	
Single-wire half-duplex communication	Х	Х
IrDA SIR ENDEC block	Х	
LIN mode	Х	
Dual clock domain and wakeup from Stop mode	Х	
Receiver timeout interrupt	Х	
Modbus communication	Х	
Auto baud rate detection	Х	
Driver Enable	Х	Х

^{1.} X = supported.

4.15 Serial peripheral interface (SPI)/Inter-integrated sound interfaces (I²S)

Up to two SPIs are able to communicate up to 18 Mbits/s in slave and master modes in full-duplex and simplex communication modes. The 3-bit prescaler gives 8 master mode frequencies and the frame size is configurable from 4 bits to 16 bits.

One standard I²S interface (multiplexed with SPI1) supporting four different audio standards can operate as master or slave at simplex communication mode. It can be configured to transfer 16 and 24 or 32 bits with16-bit or 32-bit data resolution and synchronized by a specific signal. Audio sampling frequency from 8 kHz up to 192 kHz can be set by 8-bit programmable linear prescaler. When operating in master mode it can output a clock for an external audio component at 256 times the sampling frequency.

Refer to *Table 7* for the differences between SPI1 and SPI2.

Table 7. STM32F051x SPI/I2S implementation

SPI features ⁽¹⁾	SPI1	SPI2
Hardware CRC calculation	Х	Х
Rx/Tx FIFO	Х	Х
NSS pulse mode	Х	Х
I2S mode	Х	
TI mode	Х	Х

^{1.} X = supported.

4.16 High-definition multimedia interface (HDMI) - consumer electronics control (CEC)

The device embeds a HDMI-CEC controller that provides hardware support for the Consumer Electronics Control (CEC) protocol (Supplement 1 to the HDMI standard).

This protocol provides high-level control functions between all audiovisual products in an environment. It is specified to operate at low speeds with minimum processing and memory overhead. It has a clock domain independent from the CPU clock, allowing the HDMI_CEC controller to wakeup the MCU from Stop mode on data reception.

4.17 General-purpose inputs/outputs (GPIOs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain), as input (with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions.

The I/O configuration can be locked if needed following a specific sequence in order to avoid spurious writing to the I/Os registers.

4.18 Touch sensing controller (TSC)

The STM32F051xx devices provide a simple solution for adding capacitive sensing functionality to any application. Capacitive sensing technology is able to detect the presence of a finger near an electrode which is protected from direct touch by a dielectric (glass, plastic, ...). The capacitive variation introduced by the finger (or any conductive object) is measured using a proven implementation based on a surface charge transfer acquisition principle. It consists of charging the electrode capacitance and then transferring a part of the accumulated charges into a sampling capacitor until the voltage across this capacitor has reached a specific threshold. To limit the CPU bandwidth usage this acquisition is directly managed by the hardware touch sensing controller and only requires few external components to operate. The STM32F051xx devices offer up to 18 capacitive sensing channels distributed over 6 analog I/O groups.

Table 8. Capacitive sensing GPIOs available on STM32F051x devices

Group	Capacitive sensing signal name	Pin name
	TSC_G1_IO1	PA0
1	TSC_G1_IO2	PA1
ı	TSC_G1_IO3	PA2
	TSC_G1_IO4	PA3
	TSC_G2_IO1	PA4
2	TSC_G2_IO2	PA5
۷	TSC_G2_IO3	PA6
	TSC_G2_IO4	PA7
	TSC_G3_IO1	PC5
3	TSC_G3_IO2	PB0
3	TSC_G3_IO3	PB1
	TSC_G3_IO4	PB2

Group	Capacitive sensing signal name	Pin name
	TSC_G4_IO1	PA9
4	TSC_G4_IO2	PA10
4	TSC_G4_IO3	PA11
	TSC_G4_IO4	PA12
	TSC_G5_IO1	PB3
5	TSC_G5_IO2	PB4
5	TSC_G5_IO3	PB6
	TSC_G5_IO4	PB7
	TSC_G6_IO1	PB11
6	TSC_G6_IO2	PB12
0	TSC_G6_IO3	PB13
	TSC_G6_IO4	PB14

Table 9. No. of capacitive sensing channels available on STM32F051xx devices

Analog I/O group	Number of capacitive sensing channels							
Analog I/O group	STM32F051Rx	STM32F051Cx	STM32F051Kx					
G1	3	3	3					
G2	3	3	3					
G3	3	2	2					
G4	3	3	3					
G5	3	3	3					
G6	3	3	0					
Number of capacitive sensing channels	18	17	14					

4.19 Analog to digital converter (ADC)

The 12-bit analog to digital converter has up to 16 external and 3 internal (temperature sensor, voltage reference, VBAT voltage measurement) channels and performs conversions in single-shot or scan modes. In scan mode, automatic conversion is performed on a selected group of analog inputs.

The ADC can be served by the DMA controller.

An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all selected channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

4.19.1 Temperature sensor

The temperature sensor (TS) generates a voltage V_{SENSE} that varies linearly with temperature.

The temperature sensor is internally connected to the ADC_IN16 input channel which is used to convert the sensor output voltage into a digital value.

The sensor provides good linearity but it has to be calibrated to obtain good overall accuracy of the temperature measurement. As the offset of the temperature sensor varies from chip to chip due to process variation, the uncalibrated internal temperature sensor is suitable for applications that detect temperature changes only.

To improve the accuracy of the temperature sensor measurement, each device is individually factory-calibrated by ST. The temperature sensor factory calibration data are stored by ST in the system memory area, accessible in read-only mode.

Calibration value name	Description	Memory address						
TS_CAL1	TS ADC raw data acquired at temperature of 30 °C, V _{DDA} = 3.3 V	0x1FFF F7B8 - 0x1FFF F7B9						
TS_CAL2	TS ADC raw data acquired at temperature of 110 °C	0x1FFF F7C2 - 0x1FFF F7C3						

V_{DDA}= 3.3 V

Table 10. Temperature sensor calibration values

4.19.2 Internal voltage reference (V_{REFINT})

The internal voltage reference (V_{REFINT}) provides a stable (bandgap) voltage output for the ADC and Comparators. V_{REFINT} is internally connected to the ADC_IN17 input channel. The precise voltage of V_{REFINT} is individually measured for each part by ST during production test and stored in the system memory area. It is accessible in read-only mode.

Table 11. Temperature sensor calibration values

Calibration value name	Description	Memory address
	Raw data acquired at temperature of 30 °C V _{DDA} = 3.3 V	0x1FFF F7BA - 0x1FFF F7BB

4.19.3 V_{BAT} battery voltage monitoring

This embedded hardware feature allows the application to measure the V_{BAT} battery voltage using the internal ADC channel ADC_IN18. As the V_{BAT} voltage may be higher than V_{DDA} , and thus outside the ADC input range, the V_{BAT} pin is internally connected to a bridge divider by 2. As a consequence, the converted digital value is half the V_{BAT} voltage.

4.20 Digital-to-analog converter (DAC)

The 12-bit buffered DAC channel can be used to convert digital signals into analog voltage signal outputs. The chosen design structure is composed of integrated resistor strings and an amplifier in non-inverting configuration.

This digital Interface supports the following features:

- Left or right data alignment in 12-bit mode
- Synchronized update capability
- DMA capability
- External triggers for conversion

Five DAC trigger inputs are used in the device. The DAC is triggered through the timer trigger outputs and the DAC interface is generating it's own DMA requests.

4.21 Fast low power comparators (COMP)

The device embeds two fast rail-to-rail comparators with programmable reference voltage (internal or external), hysteresis and speed (low speed for low power) and with selectable output polarity.

The reference voltage can be one of the following:

- External I/O
- DAC output pin
- Internal reference voltage or submultiple (1/4, 1/2, 3/4). Refer to *Table 24: Embedded internal reference voltage* for the value and precision of the internal reference voltage.

Both comparators can wake up from STOP mode, generate interrupts and breaks for the timers and can be also combined into a window comparator.

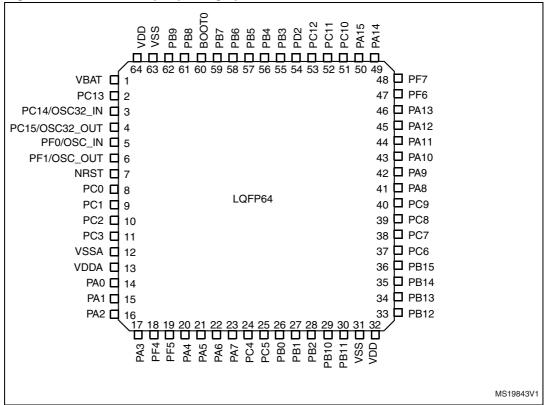
The internal voltage reference is also connected to ADC_IN17 input channel of the ADC.

4.22 Serial wire debug port (SW-DP)

An ARM SW-DP interface is provided to allow a serial wire debugging tool to be connected to the MCU.

5 Pinouts and pin description





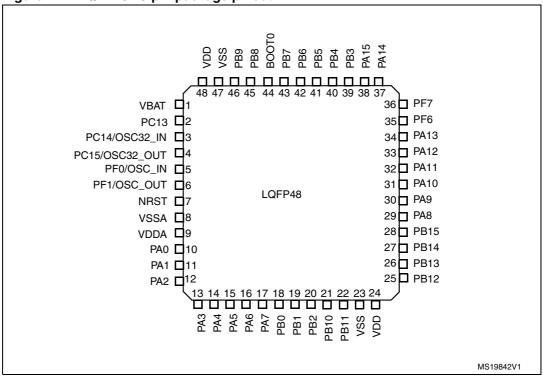
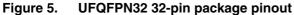
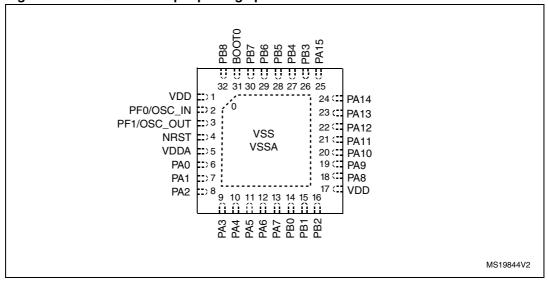


Figure 4. LQFP48 48-pin package pinout





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Table 12. Legend/abbreviations used in the pinout table

Na	me	Abbreviation	Definition		
Pin r	name	Unless otherwise specified in brackets below the pin name, the pin function during and after reset is the same as the actual pin name			
		S	Supply pin		
Pin	type	I	Input only pin		
		I/O	Input / output pin		
		FT	5 V tolerant I/O		
		FTf	5 V tolerant I/O, FM+ capable		
I/O etr	ucture	TTa 3.3 V tolerant I/O directly connected to ADC			
1/0 511	ucture	TC	Standard 3.3V I/O		
		В	Dedicated BOOT0 pin		
		RST	Bidirectional reset pin with embedded weak pull-up resistor		
No	tes	Unless otherwise specified by a note, all I/Os are set as floating inputs during and after reset			
B:	Alternate functions	Functions selected through GPIOx_AFR registers			
Pin functions	Additional functions	Functions direct	ly selected/enabled through peripheral registers		

Table 13. Pin definitions

Pin	num	ber			re		Pin functions	
LQFP64	LQFP48	UFQFPN32	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
1	1		VBAT	S			Backup power s	supply
2	2		PC13	I/O	тс	(1)(2)		RTC_TAMP1, RTC_TS, RTC_OUT, WKUP2
3	3		PC14-OSC32_IN (PC14)	I/O	тс	(1)(2)		OSC32_IN
4	4		PC15- OSC32_OUT (PC15)	I/O	тс	(1)(2)		OSC32_OUT
5	5	2	PF0-OSC_IN (PF0)	I/O	FT			OSC_IN
6	6	3	PF1-OSC_OUT (PF1)	I/O	FT			OSC_OUT
7	7	4	NRST	I/O	RST		Device reset input / internal res	et output (active low)

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Table 13. Pin definitions (continued)

Pin	num	ber			ē		Pin function	ons
LQFP64	LQFP48	UFQFPN32	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
8			PC0	I/O	TTa		EVENTOUT	ADC_IN10
9			PC1	I/O	TTa		EVENTOUT	ADC_IN11
10			PC2	I/O	TTa		EVENTOUT	ADC_IN12
11			PC3	I/O	TTa		EVENTOUT	ADC_IN13
12	8	0	VSSA	S			Analog grou	ınd
13	9	5	VDDA	S			Analog power	supply
14	10	6	PA0	I/O	TTa		USART2_CTS, TIM2_CH1_ETR, COMP1_OUT, TSC_G1_IO1	ADC_IN0, COMP1_INM6, RTC_TAMP2, WKUP1
15	11	7	PA1	I/O	TTa		USART2_RTS, TIM2_CH2, TSC_G1_IO2, EVENTOUT	ADC_IN1, COMP1_INP
16	12	8	PA2	I/O	ТТа		USART2_TX, TIM2_CH3, TIM15_CH1, COMP2_OUT, TSC_G1_IO3	ADC_IN2, COMP2_INM6
17	13	9	PA3	I/O	TTa		USART2_RX, TIM2_CH4, TIM15_CH2, TSC_G1_IO4	ADC_IN3, COMP2_INP
18			PF4	I/O	FT		EVENTOUT	
19			PF5	I/O	FT		EVENTOUT	
20	14	10	PA4	I/O	ТТа		SPI1_NSS/I2S1_WS, USART2_CK, TIM14_CH1, TSC_G2_IO1	ADC_IN4, COMP1_INM4, COMP2_INM4, DAC1_OUT
21	15	11	PA5	I/O	ТТа		SPI1_SCK/I2S1_CK, CEC, TIM2_CH_ETR, TSC_G2_IO2	ADC_IN5, COMP1_INM5, COMP2_INM5
22	16	12	PA6	I/O	TTa		SPI1_MISO/I2S1_MCK, TIM3_CH1, TIM1_BKIN, TIM16_CH1, COMP1_OUT, TSC_G2_IO3, EVENTOUT	ADC_IN6
23	17	13	PA7	I/O	TTa		SPI1_MOSI/I2S1_SD, TIM3_CH2, TIM14_CH1, TIM1_CH1N, TIM17_CH1, COMP2_OUT, TSC_G2_IO4, EVENTOUT	ADC_IN7
24			PC4	I/O	TTa		EVENTOUT	ADC_IN14
25			PC5	I/O	TTa		TSC_G3_IO1	ADC_IN15
26	18	14	PB0	I/O	TTa		TIM3_CH3, TIM1_CH2N, TSC_G3_IO2, EVENTOUT	ADC_IN8

Table 13. Pin definitions (continued)

	e 13. num		in definitions (co				Pin functions	
LQFP64	LQFP48	UFQFPN32	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
27	19	15	PB1	I/O	TTa		TIM3_CH4, TIM14_CH1, TIM1_CH3N, TSC_G3_IO3	ADC_IN9
28	20	16	PB2	I/O	FT		TSC_G3_IO4	
29	21		PB10	I/O	FT		I2C2_SCL, CEC, TIM2_CH3, TSC_SYNC	
30	22		PB11	I/O	FT		I2C2_SDA, TIM2_CH4, TSC_G6_IO1, EVENTOUT	
31	23	0	VSS	S			Digital grou	ind
32	24	17	VDD	S			Digital power s	supply
33	25		PB12	I/O	FT		SPI2_NSS, TIM1_BKIN, TSC_G6_IO2, EVENTOUT	
34	26		PB13	I/O	FT		SPI2_SCK, TIM1_CH1N, TSC_G6_IO3	
35	27		PB14	I/O	FT		SPI2_MISO, TIM1_CH2N, TIM15_CH1, TSC_G6_IO4	
36	28		PB15	I/O	FT		SPI2_MOSI, TIM1_CH3N, TIM15_CH1N, TIM15_CH2	RTC_REFIN
37			PC6	I/O	FT		TIM3_CH1	
38			PC7	I/O	FT		TIM3_CH2	
39			PC8	I/O	FT		TIM3_CH3	
40			PC9	I/O	FT		TIM3_CH4	
41	29	18	PA8	I/O	FT		USART1_CK, TIM1_CH1, EVENTOUT, MCO	
42	30	19	PA9	I/O	FT		USART1_TX, TIM1_CH2, TIM15_BKIN, TSC_G4_IO1	
43	31	20	PA10	I/O	FT		USART1_RX, TIM1_CH3, TIM17_BKIN, TSC_G4_IO2	
44	32	21	PA11	I/O	FT		USART1_CTS, TIM1_CH4, COMP1_OUT, TSC_G4_IO3, EVENTOUT	
45	33	22	PA12	I/O	FT		USART1_RTS, TIM1_ETR, COMP2_OUT, TSC_G4_IO4, EVENTOUT	
46	34	23	PA13 (SWDAT)	I/O	FT	(3)	IR_OUT, SWDAT	
47	35		PF6	I/O	FT		I2C2_SCL	
48	36		PF7	I/O	FT		I2C2_SDA	

Table 13. Pin definitions (continued)

Pin	num	ber		<u>5</u>		Pin function	ons	
LQFP64	LQFP48	UFQFPN32	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
49	37	24	PA14 (SWCLK)	I/O	FT	(3)	USART2_TX, SWCLK	
50	38	25	PA15	I/O	FT		SPI1_NSS/I2S1_WS, USART2_RX, TIM2_CH_ETR, EVENTOUT	
51			PC10	I/O	FT			
52			PC11	I/O	FT			
53			PC12	I/O	FT			
54			PD2	I/O	FT		TIM3_ETR	
55	39	26	PB3	I/O	FT		SPI1_SCK/I2S1_CK, TIM2_CH2, TSC_G5_IO1, EVENTOUT	
56	40	27	PB4	I/O	FT		SPI1_MISO/I2S1_MCK, TIM3_CH1, TSC_G5_IO2, EVENTOUT	
57	41	28	PB5	I/O	FT		SPI1_MOSI/I2S1_SD, I2C1_SMBA, TIM16_BKIN, TIM3_CH2	
58	42	29	PB6	I/O	FTf		I2C1_SCL, USART1_TX, TIM16_CH1N, TSC_G5_IO3	
59	43	30	PB7	I/O	FTf		I2C1_SDA, USART1_RX, TIM17_CH1N, TSC_G5_IO4	
60	44	31	воото	I	В		Boot memory se	election
61	45	32	PB8	I/O	FTf		I2C1_SCL, CEC, TIM16_CH1, TSC_SYNC	
62	46		PB9	I/O	FTf		I2C1_SDA, IR_OUT, TIM17_CH1, EVENTOUT	
63	47	0	VSS	S			Digital grou	nd
64	48	1	VDD	S			Digital power supply	

PC13, PC14 and PC15 are supplied through the power switch. Since the switch only sinks a limited amount of current (3 mA), the use of GPIO PC13 to PC15 in output mode is limited:

- The speed should not exceed 2 MHz with a maximum load of 30 pF

- these GPIOs must not be used as a current sources (e.g. to drive an LED).

^{2.} After the first backup domain power-up, PC13, PC14 and PC15 operate as GPIOs. Their function then depends on the content of the Backup registers which is not reset by the main reset. For details on how to manage these GPIOs, refer to the Battery backup domain and BKP register description sections in the reference manual.

^{3.} After reset, these pins are configured as SWDAT and SWCLK alternate functions, and the internal pull-up on SWDAT pin and internal pull-down on SWCLK pin are activated.



Table 14. Alternate functions selected through GPIOA_AFR registers for port A

Pin name	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
PA0		USART2_CTS	TIM2_CH1_ ETR	TSC_G1_IO1				COMP1_OUT
PA1	EVENTOUT	USART2_RTS	TIM2_CH2	TSC_G1_IO2				
PA2	TIM15_CH1	USART2_TX	TIM2_CH3	TSC_G1_IO3				COMP2_OUT
PA3	TIM15_CH2	USART2_RX	TIM2_CH4	TSC_G1_IO4				
PA4	SPI1_NSS/ I2S1_WS	USART2_CK		TSC_G2_IO1	TIM14_CH1			
PA5	SPI1_SCK/ I2S1_CK	CEC	TIM2_CH1_ ETR	TSC_G2_IO2				
PA6	SPI1_MISO/ I2S1_MCK	TIM3_CH1	TIM1_BKIN	TSC_G2_IO3		TIM16_CH1	EVENTOUT	COMP1_OUT
PA7	SPI1_MOSI/ I2S1_SD	TIM3_CH2	TIM1_CH1N	TSC_G2_IO4	TIM14_CH1	TIM17_CH1	EVENTOUT	COMP2_OUT
PA8	MCO	USART1_CK	TIM1_CH1	EVENTOUT				
PA9	TIM15_BKIN	USART1_TX	TIM1_CH2	TSC_G4_IO1				
PA10	TIM17_BKIN	USART1_RX	TIM1_CH3	TSC_G4_IO2				
PA11	EVENTOUT	USART1_CTS	TIM1_CH4	TSC_G4_IO3				COMP1_OUT
PA12	EVENTOUT	USART1_RTS	TIM1_ETR	TSC_G4_IO4				COMP2_OUT
PA13	SWDAT	IR_OUT						
PA14	SWCLK	USART2_TX						
PA15	SPI1_NSS/ I2S1_WS	USART2_RX	TIM2_CH1_ ETR	EVENTOUT				

Table 15. Alternate functions selected through GPIOB_AFR registers for port B

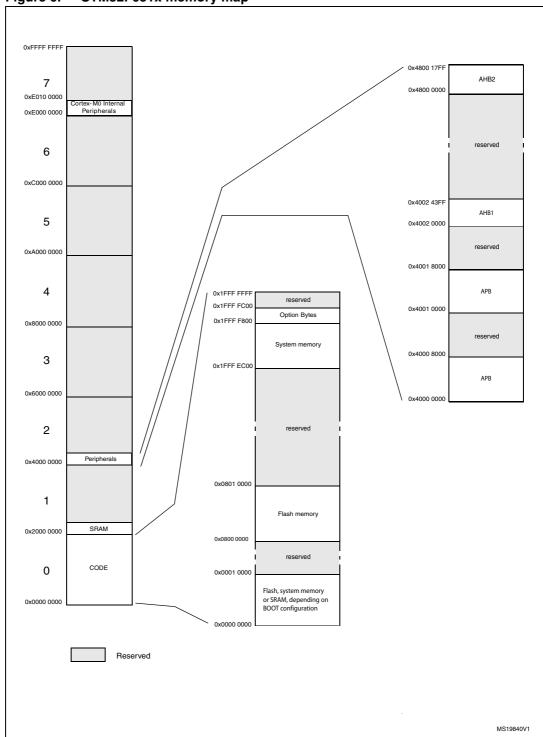
Pin name	AF0	AF1	AF2	AF3
PB0	EVENTOUT	TIM3_CH3	TIM1_CH2N	TSC_G3_IO2
PB1	TIM14_CH1	TIM3_CH4	TIM1_CH3N	TSC_G3_IO3
PB2				TSC_G3_IO4
PB3	SPI1_SCK/I2S1_CK	EVENTOUT	TIM2_CH2	TSC_G5_IO1
PB4	SPI1_MISO/I2S1_MCK	TIM3_CH1	EVENTOUT	TSC_G5_IO2
PB5	SPI1_MOSI/I2S1_SD	TIM3_CH2	TIM16_BKIN	I2C1_SMBA
PB6	USART1_TX	I2C1_SCL	TIM16_CH1N	TSC_G5_IO3
PB7	USART1_RX	I2C1_SDA	TIM17_CH1N	TSC_G5_IO4
PB8	CEC	I2C1_SCL	TIM16_CH1	TSC_SYNC
PB9	IR_OUT	I2C1_SDA	TIM17_CH1	EVENTOUT
PB10	CEC	I2C2_SCL	TIM2_CH3	TSC_SYNC
PB11	EVENTOUT	I2C2_SDA	TIM2_CH4	TSC_G6_IO1
PB12	SPI2_NSS	EVENTOUT	TIM1_BKIN	TSC_G6_IO2
PB13	SPI2_SCK		TIM1_CH1N	TSC_G6_IO3
PB14	SPI2_MISO	TIM15_CH1	TIM1_CH2N	TSC_G6_IO4
PB15	SPI2_MOSI	TIM15_CH2	TIM1_CH3N	TIM15_CH1N



STM32F051x Memory mapping

6 Memory mapping





Memory mapping STM32F051x

Table 16. STM32F051x peripheral register boundary addresses

Bus	Boundary address	Size	Peripheral	
	0x4800 1800 - 0x5FFF FFFF	~384 MB	Reserved	
	0x4800 1400 - 0x4800 17FF	1KB	GPIOF	
	0x4800 1000 - 0x4800 13FF	1KB	Reserved	
AHB2	0x4800 0C00 - 0x4800 0FFF	1KB	GPIOD	
ANDZ	0x4800 0800 - 0x4800 0BFF	1KB	GPIOC	
	0x4800 0400 - 0x4800 07FF	1KB	GPIOB	
	0x4800 0000 - 0x4800 03FF	1KB	GPIOA	
	0x4002 4400 - 0x47FF FFFF	~128 MB	Reserved	
	0x4002 4000 - 0x4002 43FF	1KB	TSC	
	0x4002 3400 - 0x4002 3FFF	3КВ	Reserved	
	0x4002 3000 - 0x4002 33FF	1KB	CRC	
	0x4002 2400 - 0x4002 2FFF	3КВ	Reserved	
AHB1	0x4002 2000 - 0x4002 23FF	1KB	FLASH Interface	
	0x4002 1400 - 0x4002 1FFF	3КВ	Reserved	
	0x4002 1000 - 0x4002 13FF	1KB	RCC	
	0x4002 0400 - 0x4002 0FFF	3КВ	Reserved	
	0x4002 0000 - 0x4002 03FF	1KB	DMA	
	0x4001 8000 - 0x4001 FFFF	32KB	Reserved	
	0x4001 5C00 - 0x4001 7FFF	9KB	Reserved	
	0x4001 5800 - 0x4001 5BFF	1KB	DBGMCU	
	0x4001 4C00 - 0x4001 57FF	3КВ	Reserved	
	0x4001 4800 - 0x4001 4BFF	1KB	TIM17	
	0x4001 4400 - 0x4001 47FF	1KB	TIM16	
	0x4001 4000 - 0x4001 43FF	1KB	TIM15	
	0x4001 3C00 - 0x4001 3FFF	1KB	Reserved	
ADD	0x4001 3800 - 0x4001 3BFF	1KB	USART1	
APB	0x4001 3400 - 0x4001 37FF	1KB	Reserved	
	0x4001 3000 - 0x4001 33FF	1KB	SPI1/I2S1	
	0x4001 2C00 - 0x4001 2FFF	1KB	TIM1	
	0x4001 2800 - 0x4001 2BFF	1KB	Reserved	
	0x4001 2400 - 0x4001 27FF	1KB	ADC	
	0x4001 0800 - 0x4001 23FF	7KB	Reserved	
	0x4001 0400 - 0x4001 07FF	1KB	EXTI	
	0x4001 0000 - 0x4001 03FF	1KB	SYSCFG + COMP	
	0x4000 8000 - 0x4000 FFFF	32KB	Reserved	

STM32F051x Memory mapping

Table 16. STM32F051x peripheral register boundary addresses (continued)

Bus	Boundary address	Size	Peripheral
	0x4000 7C00 - 0x4000 7FFF	1KB	Reserved
	0x4000 7800 - 0x4000 7BFF	1KB	CEC
	0x4000 7400 - 0x4000 77FF	1KB	DAC
	0x4000 7000 - 0x4000 73FF	1KB	PWR
	0x4000 5C00 - 0x4000 6FFF	5KB	Reserved
	0x4000 5800 - 0x4000 5BFF	1KB	I2C2
	0x4000 5400 - 0x4000 57FF	1KB	I2C1
	0x4000 4800 - 0x4000 53FF	3 KB	Reserved
	0x4000 4400 - 0x4000 47FF	1KB	USART2
	0x4000 3C00 - 0x4000 43FF	2KB	Reserved
APB	0x4000 3800 - 0x4000 3BFF	1KB	SPI2
AID	0x4000 3400 - 0x4000 37FF	1KB	Reserved
	0x4000 3000 - 0x4000 33FF	1KB	IWDG
	0x4000 2C00 - 0x4000 2FFF	1KB	WWDG
	0x4000 2800 - 0x4000 2BFF	1KB	RTC
	0x4000 2400 - 0x4000 27FF	1KB	Reserved
	0x4000 2000 - 0x4000 23FF	1KB	TIM14
	0x4000 1400 - 0x4000 1FFF	3КВ	Reserved
	0x4000 1000 - 0x4000 13FF	1KB	TIM6
	0x4000 0800 - 0x4000 0FFF	2KB	Reserved
	0x4000 0400 - 0x4000 07FF	1KB	TIM3
	0x4000 0000 - 0x4000 03FF	1KB	TIM2

7 Electrical characteristics

7.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to V_{SS}.

7.1.1 Minimum and maximum values

Unless otherwise specified, the minimum and maximum values are guaranteed in the worst conditions of ambient temperature, supply voltage and frequencies by tests in production on 100% of the devices with an ambient temperature at $T_A = 25$ °C and $T_A = T_A$ max (given by the selected temperature range).

Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes and are not tested in production. Based on characterization, the minimum and maximum values refer to sample tests and represent the mean value plus or minus three times the standard deviation (mean $\pm 3\Sigma$).

7.1.2 Typical values

Unless otherwise specified, typical data are based on $T_A = 25$ °C, $V_{DD} = V_{DDA} = 3.3$ V. They are given only as design guidelines and are not tested.

Typical ADC accuracy values are determined by characterization of a batch of samples from a standard diffusion lot over the full temperature range, where 95% of the devices have an error less than or equal to the value indicated (mean $\pm 2\Sigma$).

7.1.3 Typical curves

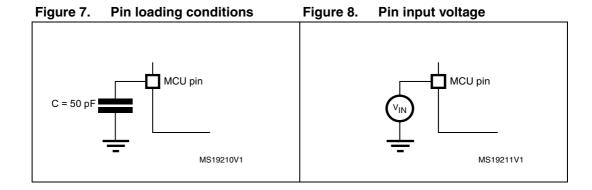
Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

7.1.4 Loading capacitor

The loading conditions used for pin parameter measurement are shown in *Figure 7*.

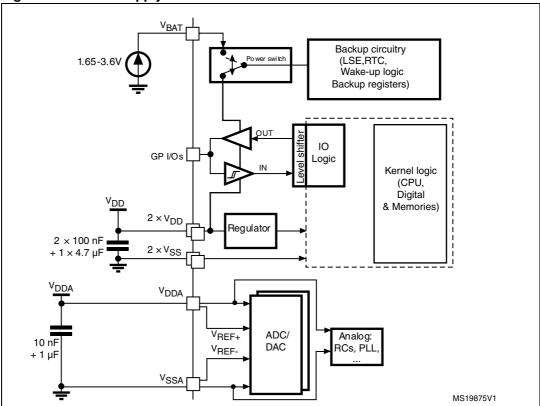
7.1.5 Pin input voltage

The input voltage measurement on a pin of the device is described in *Figure 8*.



7.1.6 Power supply scheme

Figure 9. Power supply scheme

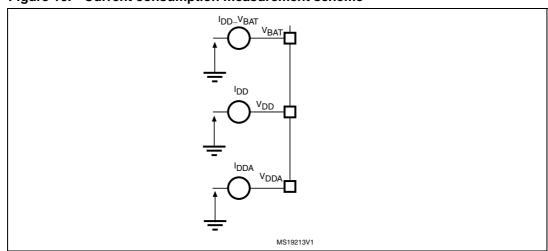


Caution:

Each power supply pair (V_{DD}/V_{SS} , V_{DDA}/V_{SSA} etc..) must be decoupled with filtering ceramic capacitors as shown above. These capacitors must be placed as close as possible to, or below, the appropriate pins on the underside of the PCB to ensure the good functionality of the device.

7.1.7 Current consumption measurement

Figure 10. Current consumption measurement scheme



7.2 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in *Table 17: Voltage characteristics*, *Table 18: Current characteristics*, and *Table 19: Thermal characteristics* may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Table 17. Voltage characteristics⁽¹⁾

Symbol	Ratings	Min	Max	Unit
V _{DD} -V _{SS}	External main supply voltage (including $V_{\rm DDA}$ and $V_{\rm DD}$)	-0.3	4.0	
V _{DD} -V _{DDA}	Allowed voltage difference for V _{DD} > V _{DDA}		0.4	v
	Input voltage on FT and FTf pins	V _{SS} - 0.3	V _{DD} + 4.0	
$V_{IN}^{(2)}$	Input voltage on TTa pins	V _{SS} – 0.3 4.0		
	Input voltage on any other pin	V _{SS} – 0.3	4.0	
IΔV _{DDx} I	Variations between different V_{DD} power pins		50	mV
IV _{SSX} – V _{SS} I	Variations between all the different ground pins		50	1110
V _{ESD(HBM)}	Electrostatic discharge voltage (human body model)	see Section 7.3. sensitivity charac		

All main power (V_{DD}, V_{DDA}) and ground (V_{SS}, V_{SSA}) pins must always be connected to the external power supply, in the permitted range.

V_{IN} maximum must always be respected. Refer to Table 18: Current characteristics for the maximum allowed injected current values.

Table 18. Current characteristics

Symbol	Ratings	Max.	Unit
I _{VDD}	Total current into V _{DD} power lines (source) ⁽¹⁾	TBD	
I _{VSS}	Total current out of V _{SS} ground lines (sink) ⁽¹⁾	TBD	
1	Output current sunk by any I/O and control pin	25	
I _{IO}	Output current source by any I/Os and control pin	- 25	mA
. (2)	Injected current on FT and FTf pins	-5 ⁽³⁾	
I _{INJ(PIN)} ⁽²⁾	Injected current on any other pin	± 5	
Σl _{INJ(PIN)}	Total injected current (sum of all I/O and control pins) ⁽⁴⁾	± 25	

- All main power (V_{DD}, V_{DDA}) and ground (V_{SS}, V_{SSA}) pins must always be connected to the external power supply, in the permitted range.
- A positive injection is induced by V_{IN}>V_{DD} while a negative injection is induced by V_{IN}<V_{SS}. I_{INJ(PIN)} must never be exceeded. Refer also to *Table 17: Voltage characteristics* for the maximum allowed input voltage values. Negative injection disturbs the analog performance of the device. See note 2 below *Table 63 on page 83*.
- 3. Positive injection is not possible on these I/Os and does not occur for input voltages lower than the specified maximum value.
- When several inputs are submitted to a current injection, the maximum ΣI_{INJ(PIN)} is the absolute sum of the
 positive and negative injected currents (instantaneous values).

Table 19. Thermal characteristics

Symbol	Ratings	Value	Unit
T _{STG}	Storage temperature range	-65 to +150	°C
T _J	Maximum junction temperature	150	°C

7.3 Operating conditions

7.3.1 General operating conditions

Table 20. General operating conditions

Symbol	Parameter	Conditions	Min	Max	Unit		
f _{HCLK}	Internal AHB clock frequency		0	48	MHz		
f _{PCLK}	Internal APB clock frequency		0	48	IVITIZ		
V _{DD}	Standard operating voltage		2	3.6	V		
V _{DDA} ⁽¹⁾	Analog operating voltage (ADC and DAC not used)	Must have a potential equal	2	3.6	V		
V _{DDA} `,	Analog operating voltage (ADC and DAC used)	to or higher than V _{DD}	2.4	3.6	v		
V _{BAT}	Backup operating voltage		1.65	3.6	V		
	Power dissipation at T _A =	LQFP64		444			
P _D	85 °C for suffix 6 or T _A =	LQFP48		364	mW		
	105 °C for suffix 7 ⁽²⁾	UFQFPN32		526			
	Ambient temperature for 6	Maximum power dissipation	-40	85	°C		
TA	suffix version	Low power dissipation ⁽³⁾	-40	105			
IA	Ambient temperature for 7	Maximum power dissipation	-40	105	°C		
	suffix version	Low power dissipation ⁽³⁾	-40	125			
TJ	Junction temperature range	6 suffix version	-40	105	°C		
IJ	Juniculon temperature range	7 suffix version	-40	125			

^{1.} When the ADC is used, refer to Table 61: ADC characteristics.

7.3.2 Operating conditions at power-up / power-down

The parameters given in *Table 21* are derived from tests performed under the ambient temperature condition summarized in *Table 20*.

Table 21. Operating conditions at power-up / power-down

Symbol	Parameter	Conditions	Min	Max	Unit
+	V _{DD} rise time rate		0	∞	
t _{VDD}	V _{DD} fall time rate		20	∞	μs/V
+	V _{DDA} rise time rate		0	8	μ5/ ν
^t VDDA	V _{DDA} fall time rate		20	8	

If T_A is lower, higher P_D values are allowed as long as T_J does not exceed T_{Jmax} (see *Table 19: Thermal characteristics*).

^{3.} In low power dissipation state, T_A can be extended to this range as long as T_J does not exceed T_{Jmax} (see *Table 19: Thermal characteristics*).

7.3.3 Embedded reset and power control block characteristics

The parameters given in *Table 22* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 20*.

Table 22. Embedded reset and power control block characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{POR/PDR} ⁽¹⁾	Power on/power down	Falling edge	1.8 ⁽²⁾	1.88	1.96	V
	reset threshold	Rising edge	1.84	1.92	2.0	V
V _{PDRhyst} ⁽¹⁾	PDR hysteresis			40		mV
t _{RSTTEMPO} ⁽³⁾	Reset temporization		1.5	2.5	4.5	ms

^{1.} The PDR detector monitors V_{DD} and also V_{DDA} (if kept enabled in the option bytes). The POR detector monitors only V_{DD} .

Table 23. Programmable voltage detector characteristics

Symbol	Parameter	Conditions	Min ⁽¹⁾	Тур	Max ⁽¹⁾	Unit
V	PVD threshold 0	Rising edge	2.1	2.18	2.26	V
V _{PVD0}	FVD tillesiloid 0	Falling edge	2	2.08	2.16	V
V	PVD threshold 1	Rising edge	2.19	2.28	2.37	V
V _{PVD1}	FVD tillesilola i	Falling edge	2.09	2.18	2.27	V
V	PVD threshold 2	Rising edge	2.28	2.38	2.48	V
V _{PVD2}	FVD tilleshold 2	Falling edge	2.18	2.28	2.38	V
V	PVD threshold 3	Rising edge	2.38	2.48	2.58	V
V _{PVD3}	FVD tillesiloid 3	Falling edge	2.28	2.38	2.48	V
V	PVD threshold 4	Rising edge	2.47	2.58	2.69	V
V _{PVD4}	PVD tilleshold 4	Falling edge	2.37	2.48	2.59	V
V	PVD threshold 5	Rising edge	2.57	2.68	2.79	V
V _{PVD5}	PVD tilleshold 5	Falling edge	2.47	2.58	2.69	V
V	PVD threshold 6	Rising edge	2.66	2.78	2.9	V
V _{PVD6}	PVD threshold 6	Falling edge	2.56	2.68	2.8	V
V	PVD threshold 7	Rising edge	2.76	2.88	3	V
V_{PVD7}	F VD tilleshold /	Falling edge	2.66	2.78	2.9	V
V _{PVDhyst} ⁽²⁾	PVD hysteresis			100		mV
I _{DD(PVD)}	PVD current consumption			0.15	0.26	μΑ

^{1.} Data based on characterization results only, not tested in production.

^{2.} The product behavior is guaranteed by design down to the minimum V_{POR/PDR} value.

^{3.} Guaranteed by design, not tested in production.

^{2.} Guaranteed by design, not tested in production.

7.3.4 Embedded reference voltage

The parameters given in *Table 24* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 20*.

Table 24. Embedded internal reference voltage

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{REFINT}	Internal reference voltage	-40 °C < T _A < +105 °C	1.16	1.2	1.25	V
	internal reference voltage	-40 °C < T _A < +85 °C	1.16	1.2	1.24 ⁽¹⁾	٧
T _{S_vrefint} ⁽²⁾	ADC sampling time when reading the internal reference voltage			5.1	17.1 ⁽³⁾	μs
V _{RERINT}	Internal reference voltage spread over the temperature range	V _{DD} = 3 V ±10 mV			10 ⁽³⁾	mV
T _{Coeff}	Temperature coefficient				100 ⁽³⁾	ppm/°C

- 1. Data based on characterization results, not tested in production.
- 2. Shortest sampling time can be determined in the application by multiple iterations.
- 3. Guaranteed by design, not tested in production.

7.3.5 Supply current characteristics

The current consumption is a function of several parameters and factors such as the operating voltage, ambient temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code.

The current consumption is measured as described in *Figure 10: Current consumption measurement scheme*.

All Run-mode current consumption measurements given in this section are performed with a reduced code that gives a consumption equivalent to CoreMark code.

Typical and maximum current consumption

The MCU is placed under the following conditions:

- All I/O pins are in input mode with a static value at V_{DD} or V_{SS} (no load)
- All peripherals are disabled except when explicitly mentioned
- The Flash memory access time is adjusted to the f_{HCLK} frequency (0 wait state from 0 to 24 MHz and 1 wait state above 24 MHz)
- Prefetch is ON when the peripherals are enabled, otherwise it is OFF (to enable prefetch the PRFTBE bit in the FLASH_ACR register must be set before clock setting and bus prescaling)
- When the peripherals are enabled f_{PCLK} = f_{HCLK}

The parameters given in *Table 25* to *Table 29* are derived from tests performed under ambient temperature and supply voltage conditions summarized in *Table 20*.

Table 25. Typical and maximum current consumption from V_{DD} supply at V_{DD} = 3.6 V

Table 25	7.	and maxim			l periph	-				erals dis		
Symbol	Parameter	Conditions	f _{HCLK}	T	M	lax @ T	A ⁽¹⁾	T	IV	lax @ T	A ⁽¹⁾	Unit
				Тур	25 °C	85 °C	105 °C	Тур	25 °C	85 °C	105 °C	
		HSE	48 MHz	22	22.8	22.8	23.8	11.8	12.7	12.7	13.3	
		bypass,	32 MHz	15	15.5	15.5	16.0	7.6	8.7	8.7	9.0	
		PLL on	24 MHz	12.2	13.2	13.2	13.6	7.2	7.9	7.9	8.1	
	Supply current in	HSE	8 MHz	4.4	5.2	5.2	5.4	2.7	2.9	2.9	3.0	
	Run mode,	bypass, PLL off	1 MHz	1	1.3	1.3	1.4	0.7	0.9	0.9	0.9	
	executing		48 MHz	22	22.8	22.8	23.8	11.8	12.7	12.7	13.3	
	from Flash	HSI clock, PLL on	32 MHz	15	15.5	15.5	16.0	7.6	8.7	8.7	9.0	
			24 MHz	12.2	13.2	13.2	13.6	7.2	7.9	7.9	8.1	
		HSI clock, PLL off	8 MHz	4.4	5.2	5.2	5.4	2.7	2.9	2.9	3.0	
		HSE	48 MHz	22.2	23.2 ⁽²⁾	23.2	24.4 ⁽²⁾	12.0	12.7 ⁽²⁾	12.7	13.3 ⁽²⁾	
		bypass,	32 MHz	15.4	16.3	16.3	16.8	7.8	8.7	8.7	9.0	
		PLL on	24 MHz	11.2	12.2	12.2	12.8	6.2	7.9	7.9	8.1	
	Supply current in	HSE	8 MHz	4.0	4.5	4.5	4.7	1.9	2.9	2.9	3.0	
I _{DD}	Run mode,	bypass,	1 MHz	0.6	0.8	0.8	0.9	0.3	0.6	0.6	0.7	mA
	executing	HSI clock, PLL on	48 MHz	22.2	23.2	23.2	24.4	12.0	12.7	12.7	13.3	
	from RAM		32 MHz	15.4	16.3	16.3	16.8	7.8	8.7	8.7	9.0	
			24 MHz	11.2	12.2	12.2	12.8	6.2	7.9	7.9	8.1	
		HSI clock, PLL off	8 MHz	4.0	4.5	4.5	4.7	1.9	2.9	2.9	3.0	
		HSE	48 MHz	14	15.3 ⁽²⁾	15.3	16.0 ⁽²⁾	2.8	3.0 ⁽²⁾	3.0	3.2 ⁽²⁾	
		bypass,	32 MHz	9.5	10.2	10.2	10.7	2.0	2.1	2.1	2.3	
	Supply	PLL on	24 MHz	7.3	7.8	7.8	8.3	1.5	1.7	1.7	1.9	
	current in	HSE	8 MHz	2.6	2.9	2.9	3.0	0.6	8.0	0.8	0.8	
	Sleep mode, code executing	bypass, PLL off	1 MHz	0.4	0.6	0.6	0.6	0.2	0.4	0.4	0.4	
			48 MHz	14	15.3	15.3	16.0	3.8	4.0	4.1	4.2	
	from Flash or RAM	HSI clock, PLL on	32 MHz	9.5	10.2	10.2	10.7	2.6	2.7	2.8	2.8	
	0. 1 t/ (IV)		24 MHz	7.3	7.8	7.8	8.3	2.0	2.1	2.1	2.1	
		HSI clock, PLL off	8 MHz	2.6	2.9	2.9	3.0	0.6	0.8	0.8	0.8	

^{1.} Data based on characterization results, not tested in production unless otherwise specified.

^{2.} Data based on characterization results and tested in production with code executing from RAM.

Table 26. Typical and maximum current consumption from the V_{DDA} supply

					V _{DDA}	$V_{DDA} = 2.4 V$			V _{DDA}	= 3.6 V	,	
Symbol	Parameter	Conditions (1)	f _{HCLK}	Tun	М	Лах @ Т _А ⁽²⁾		Tun	Max @ T _A ⁽²⁾			Unit
				Тур	25 °C	85 °C	105 °C	Тур	25 °C	85 °C	105 °C	
		HSE	48 MHz	150	170	178	182	164	183	195	198	
		bypass,	32 MHz	104	121	126	128	113	129	135	138	
	Cumply	PLL on	24 MHz	82	96	100	103	88	102	106	108	
	Supply current in	HSE	8 MHz	2.0	2.7	3.1	3.3	3.5	3.8	4.1	4.4	
	Run mode, code	bypass, PLL off	1 MHz	2.0	2.7	3.1	3.3	3.5	3.8	4.1	4.4	
	executing from Flash		48 MHz	220	240	248	252	244	263	275	278	
	or RAM	HSI clock, PLL on	32 MHz	174	191	196	198	193	209	215	218	
			24 MHz	152	167	173	174	168	183	190	192	
		HSI clock, PLL off	8 MHz	72	79	82	83	83.5	91	94	95	
I _{DDA}		HSE	48 MHz	150	170	178	182	164	183	195	198	μA
		bypass,	32 MHz	104	121	126	128	113	129	135	138	
	Supply	PLL on	24 MHz	82	96	100	103	88	102	106	108	
	current in	HSE	8 MHz	2.0	2.7	3.1	3.3	3.5	3.8	4.1	4.4	
	Sleep mode,	bypass, PLL off	1 MHz	2.0	2.7	3.1	3.3	3.5	3.8	4.1	4.4	
	code executing from Flash or RAM		48 MHz	220	240	248	252	244	263	275	278	
		Flash HSI clock,	32 MHz	174	191	196	198	193	209	215	218	
	OI I TAIVI		24 MHz	152	167	173	174	168	183	190	192	
		HSI clock, PLL off	8 MHz	72	79	82	83	83.5	91	94	95	

Current consumption from the V_{DDA} supply is independent of whether the peripherals are on or off. Furthermore when the PLL is off, I_{DDA} is independent from the frequency.

^{2.} Data based on characterization results, not tested in production.

Table 27. Typical and maximum V_{DD} consumption in Stop and Standby modes

Symbol Pa		Conditions	Typ @V _{DD} (V _{DD} = V _{DDA})							Max ⁽¹⁾			
	Parameter		2.0 V	2.4 V	2.7 V	3.0 V	3.3 V	3.6 V	T _A = 25 °C	T _A = 85 °C	T _A = 105 °C	Unit	
Cu Sto I _{DD} Su cu Sto	Supply	Regulator in run mode, all oscillators OFF	15	15.1	15.25	15.45	15.7	16	22 ⁽²⁾	48	64 ⁽²⁾		
	current in Stop mode	Regulator in low-power mode, all oscillators OFF	3.15	3.25	3.35	3.45	3.7	4	7 ⁽²⁾	32	45 ⁽²⁾	μΑ	
	Supply	LSI ON and IWDG ON	0.8	0.95	1.05	1.2	1.35	1.5					
	current in Standby mode	LSI OFF and IWDG OFF	0.65	0.75	0.85	0.95	1.1	1.3	2 ⁽²⁾	2.5	3 ⁽²⁾		

^{1.} Data based on characterization results, not tested in production unless otherwise specified.

^{2.} Data based on characterization results and tested in production.

Table 28. Typical and maximum V_{DDA} consumption in Stop and Standby modes

					Тур @	V _{DD} (V _{DD} =	V _{DDA})			Max ⁽¹⁾			
Symbol	Parameter		Conditions	2.0 V	2.4 V	2.7 V	3.0 V	3.3 V	3.6 V	T _A = 25 °C	T _A = 85 °C	T _A = 105 °C	Unit	
	Supply current in	NO	Regulator in run mode, all oscillators OFF	1.85	2	2.15	2.3	2.45	2.6	3.5	3.5	4.5		
	Stop mode	υg	Regulator in low-power mode, all oscillators OFF	1.85	2	2.15	2.3	2.45	2.6	3.5	3.5	4.5		
	Supply		LSI ON and IWDG ON	2.25	2.5	2.65	2.85	3.05	3.3					
	current in Standby mode	V _{DDA}	LSI OFF and IWDG OFF	1.75	1.9	2	2.15	2.3	2.5	3.5	3.5	4.5		
I _{DDA}	Supply	Supply 出	Regulator in run mode, all oscillators OFF	1.11	1.15	1.18	1.22	1.27	1.35				μΑ	
	current in Stop mode	orin	Regulator in low-power mode, all oscillators OFF	1.11	1.15	1.18	1.22	1.27	1.35					
			_	LSI ON and IWDG ON	1.5	1.58	1.65	1.78	1.91	2.04				
	current in Standby mode	V _{DDA}	LSI OFF and IWDG OFF	1	1.02	1.05	1.05	1.15	1.22				μА	

^{1.} Data based on characterization results, not tested in production.

Table 29. Typical and maximum current consumption from V_{BAT} supply

					Тур @	V _{BAT}	•			Max ⁽¹⁾		
Symbol	Parameter	Conditions	= 1.65 V	= 1.8 V	= 2.4 V	= 2.7 V	= 3.3 V	= 3.6 V	T _A = 25 °C	T _A = 85 °C	T _A = 105 °C	Unit
I _{DD} _vbat	Backup domain	LSE & RTC ON; "Xtal mode": lower driving capability; LSEDRV[1:0] = '00'	0.41	0.43	0.53	0.58	0.71	0.80	0.85	1.1	1.5	Unit
	supply current	LSE & RTC ON; "Xtal mode" higher driving capability; LSEDRV[1:0] = '11'	0.71	0.75	0.85	0.91	1.06	1.16	1.25	1.55	2	μΑ

^{1.} Data based on characterization results, not tested in production.

Typical current consumption

The MCU is placed under the following conditions:

- V_{DD}=V_{DDA}=3.3 V
- All I/O pins are in analog input configuration
- The Flash access time is adjusted to f_{HCLK} frequency (0 wait states from 0 to 24 MHz, 1 wait state above)
- Prefetch is ON when the peripherals are enabled, otherwise it is OFF
- When the peripherals are enabled, f_{PCLK} = f_{HCLK}
- PLL is used for frequencies greater than 8 MHz
- AHB prescaler of 2, 4, 8 and 16 is used for the frequencies 4 MHz, 2 MHz, 1 MHz and 500 kHz respectively
- $\bullet~$ A development tool is connected to the board and the parasitic pull-up current is around 30 μA

Table 30. Typical current consumption in Run mode, code with data processing running from Flash

				Ту	/p	
Symbol	Parameter	Conditions	f _{HCLK}	Peripherals enabled	Peripherals disabled	Unit
			48 MHz	23.3	11.5	
			36 MHz	17.6	9.0	
			32 MHz	15.9	8.0	
I			24 MHz	12.4	7.5	
	Supply current in Run mode from V _{DD}		16 MHz	8.5	5.2	mA
I _{DD}	supply		8 MHz	4.5	3.0	ША
			4 MHz	2.8	1.9	
			2 MHz	1.7	1.3	
		Running from HSE crystal	1 MHz	1.3	1.0	
		clock 8 MHz,	500 kHz	1.0	0.9	
		code	48 MHz	158	158	
		executing from Flash	36 MHz	120	120	
			32 MHz	108	108	
			24 MHz	83	83	
I _{DDA}	Supply current in Run mode from V _{DDA}		16 MHz	60	60	μΑ
DDA	supply		8 MHz	2.43	2.43	μΛ
			4 MHz	2.43	2.43	
			2 MHz	2.43	2.43	
			1 MHz	2.43	2.43	
			500 kHz	2.43	2.43	

Table 31. Typical current consumption in Sleep mode, code running from Flash or RAM

				Ту	/p		
Symbol	Parameter	Conditions	f _{HCLK}	Peripherals enabled	Peripherals disabled	Unit	
			48 MHz	13.9	2.98		
			36 MHz	10.55	2.84		
			32 MHz	9.6	2.6		
			24 MHz	7.23	2.09		
I _{DD}	Supply current in		16 MHz	5.01	1.58		
	Sleep mode from V _{DD}		8 MHz	2.68	0.99	mA	
	supply		4 MHz	1.81	0.85		
			2 MHz	1.27	0.77		
			1 MHz	1.03	0.73		
		Running from HSE crystal	500 kHz	0.9	0.71		
		clock 8 MHz,	125 kHz	0.78	0.69		
		code executing	48 MHz	158	157		
		from Flash or RAM	36 MHz	119	119		
			32 MHz	108	107		
			24 MHz	83	83		
	Supply current in		16 MHz	60	60		
I _{DDA}	Sleep mode from		8 MHz	2.36	2.38	μΑ	
	V _{DDA} supply		4 MHz	2.36	2.38		
			2 MHz	2.36	2.38		
			1 MHz	2.36	2.38		
			500 kHz	2.36	2.38		
			125 kHz	2.36	2.38		

I/O system current consumption

The current consumption of the I/O system has two components: static and dynamic.

I/O static current consumption

All the I/Os used as inputs with pull-up generate current consumption when the pin is externally held low. The value of this current consumption can be simply computed by using the pull-up/pull-down resistors values given in *Table 50: I/O static characteristics*.

For the output pins, any external pull-down or external load must also be considered to estimate the current consumption.

Additional I/O current consumption is due to I/Os configured as inputs if an intermediate voltage level is externally applied. This current consumption is caused by the input Schmitt trigger circuits used to discriminate the input value. Unless this specific configuration is required by the application, this supply current consumption can be avoided by configuring these I/Os in analog mode. This is notably the case of ADC input pins which should be configured as analog inputs.

Caution:

Any floating input pin can also settle to an intermediate voltage level or switch inadvertently, as a result of external electromagnetic noise. To avoid current consumption related to floating pins, they must either be configured in analog mode, or forced internally to a definite digital value. This can be done either by using pull-up/down resistors or by configuring the pins in output mode.

I/O dynamic current consumption

In addition to the internal peripheral current consumption measured previously (see *Table 33: Peripheral current consumption*), the I/Os used by an application also contribute to the current consumption. When an I/O pin switches, it uses the current from the MCU supply voltage to supply the I/O pin circuitry and to charge/discharge the capacitive load (internal or external) connected to the pin:

$$I_{SW} = V_{DD} \times f_{SW} \times C$$

where

 $I_{\mbox{\scriptsize SW}}$ is the current sunk by a switching I/O to charge/discharge the capacitive load

 $\ensuremath{V_{DD}}$ is the MCU supply voltage

f_{SW} is the I/O switching frequency

C is the total capacitance seen by the I/O pin: $C = C_{INT} + C_{FXT} + C_{S}$

C_S is the PCB board capacitance including the pad pin.

The test pin is configured in push-pull output mode and is toggled by software at a fixed frequency.

Table 32. Switching output I/O current consumption

Symbol	Parameter	Conditions ⁽¹⁾	I/O toggling frequency (f _{SW})	Тур	Unit	
			4 MHz	0.07		
		V _{DD} = 3.3 V	8 MHz	0.15		
		C =C _{INT}	16 MHz	0.31		
			24 MHz	0.53		
			48 MHz	0.92		
			4 MHz	0.18		
		V _{DD} = 3.3 Volts	8 MHz	0.37		
		C _{EXT} = 0 pF	16 MHz	0.76		
		$C = C_{INT} + C_{EXT} + C_{S}$	24 MHz	1.39		
			48 MHz	2.188		
			4 MHz	0.32	, mA	
		V _{DD} = 3.3 Volts	8 MHz	0.64		
		C _{EXT} = 10 pF	16 MHz	1.25		
		$C = C_{INT} + C_{EXT} + C_{S}$	24 MHz	2.23		
I _{SW}	I/O current		48 MHz	4.442		
iSW	consumption	$V_{DD} = 3.3 \text{ Volts}$ $C_{EXT} = 22 \text{ pF}$ $C = C_{INT} + C_{EXT} + C_{S}$	4 MHz	0.49		
			8 MHz	0.94		
			16 MHz	2.38		
		INT EXT 5	24 MHz	3.99		
			4 MHz	0.64		
		$V_{DD} = 3.3 \text{ Volts}$	8 MHz	1.25	1	
		$C_{EXT} = 33 \text{ pF}$ $C = C_{INT} + C_{EXT} + C_{S}$	16 MHz	3.24		
		INT EXT 5	24 MHz	5.02		
		V _{DD} = 3.3 Volts	4 MHz	0.81		
		C _{EXT} = 47 pF	8 MHz	1.7		
		$C = C_{INT} + C_{EXT} + C_{S}$ $C = C_{int}$	16 MHz	3.67		
		V _{DD} = 2.4 Volts	4 MHz	0.66		
		$C_{\text{EXT}} = 47 \text{ pF}$	8 MHz	1.43		
		$C = C_{INT} + C_{EXT} + C_{S}$	16 MHz	2.45		
		$C = C_{int}$	24 MHz	4.97		

^{1.} $C_S = 7 \text{ pF (estimated value)}.$

On-chip peripheral current consumption

The current consumption of the on-chip peripherals is given in *Table 33*. The MCU is placed under the following conditions:

- all I/O pins are in input mode with a static value at V_{DD} or V_{SS} (no load)
- all peripherals are disabled unless otherwise mentioned
- the given value is calculated by measuring the current consumption
 - with all peripherals clocked off
 - with only one peripheral clocked on
- ambient operating temperature and V_{DD} supply voltage conditions summarized in Table 17

Table 33. Peripheral current consumption⁽¹⁾

Peripheral	Typical consu	umption at 25 °C	Unit
renpheral	I _{DD}	I _{DDA}	01111
ADC ⁽²⁾	0.53	0.964	
CEC	0.24		
CRC	0.10		
DAC ⁽³⁾	0.27	0.408	
DBGMCU	0.18		
DMA	0.35		
GPIOA	0.48		
GPIOB	0.58		
GPIOC	0.12		
GPIOD	0.04		
GPIOF	0.06		
I2C1	0.43		
I2C2	0.42		
PWR	0.22		
SPI1/I2S1	0.63		mA
SPI2	0.53		
SYSCFG & COMP	0.28	See note ⁽⁴⁾	
TIM1	1.01		
TIM2	1.00		
TIM3	0.78		
TIM6	0.32		
TIM14	0.45		
TIM15	0.66		
TIM16	0.57		
TIM17	0.59		
TSC	0.28		
USART1	1.07		
USART2	0.48		
WWDG	0.22		

^{1.} $f_{HCLK} = 48$ MHz, $f_{PCLK} = f_{HCLK}$, default prescaler value for each peripheral.

^{2.} ADC is in ready state after setting the ADEN bit in the ADC_CR register (ADRDY bit in ADC_ISR is high).

^{3.} DAC channel 1 enabled by setting EN1 bit in DAC_CR.

^{4.} COMP I_{DDA} is specified as I_{DD(COMP)} in *Table 65: Comparator characteristics*

7.3.6 External clock source characteristics

High-speed external user clock generated from an external source

In bypass mode the HSE oscillator is switched off and the input pin is a standard GPIO.

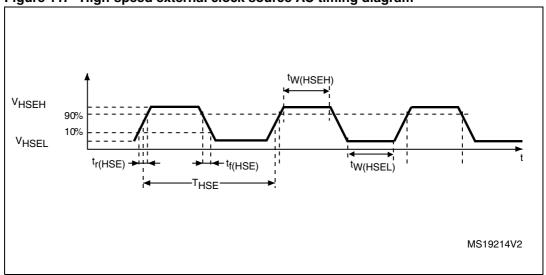
The external clock signal has to respect the I/O characteristics in *Section 7.3.13*. However, the recommended clock input waveform is shown in *Figure 11*.

Table 34. High-speed external user clock characteristics

Symbol	Parameter ⁽¹⁾	Conditions	Min	Тур	Max	Unit
f _{HSE_ext}	User external clock source frequency		1	8	32	MHz
V _{HSEH}	OSC_IN input pin high level voltage		0.7V _{DD}		V_{DD}	V
V _{HSEL}	OSC_IN input pin low level voltage		V_{SS}		0.3V _{DD}	
t _{w(HSEH)} t _{w(HSEL)}	OSC_IN high or low time		15			ns
t _{r(HSE)}	OSC_IN rise or fall time				20	113

^{1.} Guaranteed by design, not tested in production.

Figure 11. High-speed external clock source AC timing diagram



Low-speed external user clock generated from an external source

In bypass mode the LSE oscillator is switched off and the input pin is a standard GPIO.

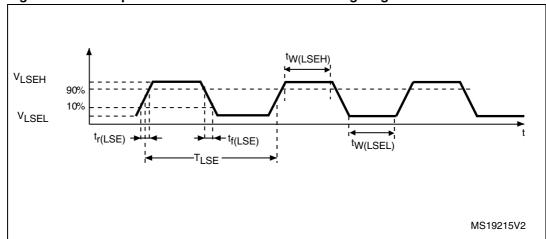
The external clock signal has to respect the I/O characteristics in *Section 7.3.13*. However, the recommended clock input waveform is shown in *Figure 12*.

Table 35. Low-speed external user clock characteristics

Symbol	Parameter ⁽¹⁾	Conditions	Min	Тур	Max	Unit
f _{LSE_ext}	User External clock source frequency			32.768	1000	kHz
V _{LSEH}	OSC32_IN input pin high level voltage		0.7V _{DD}		V_{DD}	٧
V _{LSEL}	OSC32_IN input pin low level voltage		V _{SS}		0.3V _{DD}	V
t _{w(LSEH)} t _{w(LSEL)}	OSC32_IN high or low time		450			ns
t _{r(LSE)}	OSC32_IN rise or fall time				50	115

^{1.} Guaranteed by design, not tested in production.

Figure 12. Low-speed external clock source AC timing diagram



High-speed external clock generated from a crystal/ceramic resonator

The high-speed external (HSE) clock can be supplied with a 4 to 32 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on design simulation results obtained with typical external components specified in *Table 36*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Table 36.	HSE oscillator characteristics
IADIC JU.	TISE USCHIALUI CHALACIEHSHUS

Symbol	Parameter	Conditions ⁽¹⁾	Min ⁽²⁾	Тур	Max ⁽²⁾	Unit
f _{OSC_IN}	Oscillator frequency		4	8	32	MHz
R_{F}	Feedback resistor			200		kΩ
		During startup ⁽³⁾			8.5	
		V _{DD} =3.3 V, Rm= 30Ω, CL=10 pF@8 MHz		0.4		
		V _{DD} =3.3 V, Rm= 45Ω, CL=10 pF@8 MHz		0.5		
I _{DD}	HSE current consumption	V _{DD} =3.3 V, Rm= 30Ω, CL=5 pF@32 MHz		0.8		mA
	V _{DD} =3.3 V, Rm= 30Ω, CL=10 pF@32 MHz			1		
		V _{DD} =3.3 V, Rm= 30Ω, CL=20 pF@32 MHz		1.5		
9 _m	Oscillator transconductance	Startup	10			mA/V
t _{SU(HSE)} ⁽⁴⁾	Startup time	V _{DD} is stabilized		2		ms

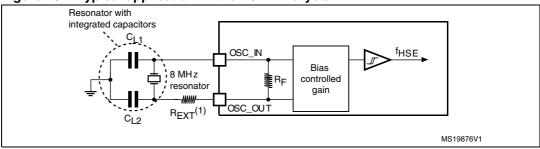
- 1. Resonator characteristics given by the crystal/ceramic resonator manufacturer.
- 2. Guaranteed by design, not tested in production.
- 3. This consumption level occurs during the first 2/3 of the $t_{SU(HSE)}$ startup time
- 4. t_{SU(HSE)} is the startup time measured from the moment it is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer

For C_{L1} and C_{L2} , it is recommended to use high-quality external ceramic capacitors in the 5 pF to 20 pF range (typ.), designed for high-frequency applications, and selected to match the requirements of the crystal or resonator (see *Figure 13*). C_{L1} and C_{L2} are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of C_{L1} and C_{L2} . PCB and MCU pin capacitance must be included (10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing C_{L1} and C_{L2} .

Note:

For information on electing the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

Figure 13. Typical application with an 8 MHz crystal



1. R_{EXT} value depends on the crystal characteristics.

Low-speed external clock generated from a crystal resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal resonator oscillator. All the information given in this paragraph are based on design simulation results obtained with typical external components specified in *Table 37*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Table 37. LSE oscillator characteristics (f_{LSE} = 32.768 kHz)

Symbol	Parameter	Conditions ⁽¹⁾	Min ⁽²⁾	Тур	Max ⁽²⁾	Unit
I _{DD}		LSEDRV[1:0]=00 lower driving capability		0.5	0.9	
	LSE current consumption	LSEDRV[1:0]= 01 medium low driving capability			1	
	Loc current consumption	LSEDRV[1:0] = 10 medium high driving capability			1.3	μА
		LSEDRV[1:0]=11 higher driving capability			1.6	
		LSEDRV[1:0]=00 lower driving capability	5			
	Oscillator	LSEDRV[1:0]= 01 medium low driving capability	8			μΑ/V
9 _m	transconductance	LSEDRV[1:0] = 10 medium high driving capability	15			μΑνν
		LSEDRV[1:0]=11 higher driving capability	25			
t _{SU(LSE)} (3)	Startup time	V _{DD} is stabilized		2		S

Refer to the note and caution paragraphs below the table, and to the application note AN2867 "Oscillator design guide for ST microcontrollers".

Note: For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

^{2.} Guaranteed by design, not tested in production.

^{3.} t_{SU(LSE)} is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768 kHz oscillation is reached. This value is measured for a standard crystal and it can vary significantly with the crystal manufacturer

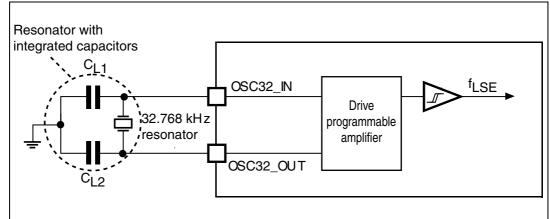


Figure 14. Typical application with a 32.768 kHz crystal

Note:

An external resistor is not required between OSC32_IN and OSC32_OUT and it is forbidden to add one.

7.3.7 Internal clock source characteristics

The parameters given in *Table 38* are derived from tests performed under ambient temperature and supply voltage conditions summarized in *Table 20*.

High-speed internal (HSI) RC oscillator

Table 38. HSI oscillator characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f _{HSI}	Frequency			8		MHz
TRIM	HSI user trimming step				1 ⁽²⁾	%
DuCy _(HSI)	Duty cycle		45 ⁽²⁾		55 ⁽²⁾	%
	Accuracy of the HSI oscillator (factory calibrated)	$T_A = -40$ to 105 °C	-2.4 ⁽³⁾		2.5 ⁽³⁾	%
400		T _A = -10 to 85 °C	-2.2 ⁽³⁾		1.3 ⁽³⁾	%
ACC _{HSI}		T _A = 0 to 70 °C	-1.9 ⁽³⁾		1.3 ⁽³⁾	%
		T _A = 25 °C	-1		1	%
t _{su(HSI)}	HSI oscillator startup time		1 ⁽²⁾		2 ⁽²⁾	μs
I _{DD(HSI)}	HSI oscillator power consumption			80	100 ⁽²⁾	μΑ

^{1.} $V_{DDA} = 3.3 \text{ V}$, $T_A = -40 \text{ to } 105 \,^{\circ}\text{C}$ unless otherwise specified.

^{2.} Guaranteed by design, not tested in production.

^{3.} Data based on characterization results, not tested in production.

High-speed internal 14 MHz (HSI14) RC oscillator (dedicated to ADC)

Table 39. HSI14 oscillator characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f _{HSI14}	Frequency			14		MHz
TRIM	HSI14 user-trimming step				1 ⁽²⁾	%
DuCy _(HSI14)	Duty cycle		45 ⁽²⁾		55 ⁽²⁾	%
		$T_A = -40 \text{ to } 105 ^{\circ}\text{C}$	TBD ⁽³⁾		TBD ⁽³⁾	%
400	Accuracy of the HSI14	$T_A = -10 \text{ to } 85 ^{\circ}\text{C}$	TBD ⁽³⁾		TBD ⁽³⁾	%
ACC _{HSI14}	oscillator (factory calibrated)	T _A = 0 to 70 °C	TBD ⁽³⁾		TBD ⁽³⁾	%
		T _A = 25 °C	-1		1	%
t _{su(HSI14)}	HSI14 oscillator startup time		1 ⁽²⁾		2 ⁽²⁾	μs
I _{DD(HSI14)}	HSI14 oscillator power consumption			100	150 ⁽²⁾	μΑ

- 1. $V_{DDA} = 3.3 \text{ V}$, $T_A = -40 \text{ to } 105 \,^{\circ}\text{C}$ unless otherwise specified.
- 2. Guaranteed by design, not tested in production.
- 3. Data based on characterization results, not tested in production.

Low-speed internal (LSI) RC oscillator

Table 40. LSI oscillator characteristics⁽¹⁾

Symbol	I Parameter		Тур	Max	Unit
f _{LSI}	Frequency	30	40	50	kHz
t _{su(LSI)} ⁽²⁾	t _{su(LSI)} ⁽²⁾ LSI oscillator startup time			85	μs
I _{DD(LSI)} ⁽²⁾	LSI oscillator power consumption		0.75	1.2	μΑ

- 1. V_{DDA} = 3.3 V, T_A = -40 to 105 °C unless otherwise specified.
- 2. Guaranteed by design, not tested in production.

Wakeup time from low-power mode

The wakeup times given in *Table 41* is measured on a wakeup phase with a 8-MHz HSI RC oscillator. The event used to wake up the device depends from the current operating mode:

- Stop or sleep mode: the wakeup event is WFE
- The wakeup pin used in stop and sleep mode is PA0 and in standby mode is the PA1.

All timings are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 20*.

Table 41. Low-power mode wakeup timings

Symbol	Parameter	Conditions	Typ @Vod				Max	Unit	
	raiailletei		= 2.4 V	= 2.7 V	= 3 V	= 3.3 V	IVIAX	Oill	
+	Wakeup from Stop	Regulator in run mode	4.2	4.2	4.2	4.2	4.2	5	
twustop	mode	Regulator in low power mode	8.05	7.05	6.6	6.27	6.05	9	
t _{WUSTANDBY} Wakeup from Standby mode		60.35	55.6	53.5	52.02	50.96		μs	
t _{WUSLEEP}	Wakeup from Sleep mode		1.1	1.1	1.1	1.1	1.1		

7.3.8 PLL characteristics

The parameters given in *Table 42* are derived from tests performed under ambient temperature and supply voltage conditions summarized in *Table 20*.

Table 42. PLL characteristics

Symbol	Parameter		Unit		
Symbol	Parameter	Min	Тур	Max	Offic
f	PLL input clock ⁽¹⁾	1 ⁽²⁾	8.0	24 ⁽²⁾	MHz
f _{PLL_IN}	PLL input clock duty cycle	40 ⁽²⁾		60 ⁽²⁾	%
f _{PLL_OUT}	PLL multiplier output clock	16 ⁽²⁾		48	MHz
t _{LOCK}	PLL lock time			200 ⁽²⁾	μs
Jitter _{PLL}	Cycle-to-cycle jitter			300 ⁽²⁾	ps

Take care to use the appropriate multiplier factors to obtain PLL input clock values compatible with the range defined by f_{PLL_OUT}.

^{2.} Guaranteed by design, not tested in production.

7.3.9 Memory characteristics

Flash memory

The characteristics are given at $T_A = -40$ to 105 $^{\circ}C$ unless otherwise specified.

Table 43. Flash memory characteristics

Symbol	Parameter	Conditions	Min	Тур	Max ⁽¹⁾	Unit
t _{prog}	16-bit programming time	$T_A = -40 \text{ to } +105 ^{\circ}\text{C}$	40	53.5	60	μs
t _{ERASE}	Page (1 KB) erase time	$T_A = -40 \text{ to } +105 ^{\circ}\text{C}$	20		40	ms
t _{ME}	Mass erase time	$T_A = -40 \text{ to } +105 ^{\circ}\text{C}$	20		40	ms
	Supply ourrent	Write mode			10	mA
IDD	Supply current	Erase mode			12	mA
V _{prog}	Programming voltage		2		3.6	V

^{1.} Guaranteed by design, not tested in production.

Table 44. Flash memory endurance and data retention

Symbol	Parameter Conditions		Value	Unit
Symbol	Parameter Conditions	Conditions	Min ⁽¹⁾	Offic
N _{END}	Endurance	$T_A = -40$ to +85 °C (6 suffix versions) $T_A = -40$ to +105 °C (7 suffix versions)	10	kcycles
		1 kcycle ⁽²⁾ at T _A = 85 °C	30	
t _{RET}	Data retention	1 kcycle ⁽²⁾ at T _A = 105 °C	10	Years
		10 kcycles ⁽²⁾ at T _A = 55 °C	20	

^{1.} Data based on characterization results, not tested in production.

^{2.} Cycling performed over the whole temperature range.

7.3.10 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports). the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- Electrostatic discharge (ESD) (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- FTB: A Burst of Fast Transient voltage (positive and negative) is applied to V_{DD} and V_{SS} through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.

A device reset allows normal operations to be resumed.

The test results are given in *Table 45*. They are based on the EMS levels and classes defined in application note AN1709.

Table 45. EMS characteristics

Symbol	Parameter	Conditions	Level/ Class
V _{FESD}	Voltage limits to be applied on any I/O pin to induce a functional disturbance	V_{DD} = 3.3 V, LQFP64, T_A = +25 °C, f_{HCLK} = 48 MHz conforms to IEC 61000-4-2	2B
V _{EFTB}	Fast transient voltage burst limits to be applied through 100 pF on V _{DD} and V _{SS} pins to induce a functional disturbance	V_{DD} = 3.3 V, LQFP64, T_A = +25 °C, f_{HCLK} = 48 MHz conforms to IEC 61000-4-4	3B

Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.

Software recommendations

The software flowchart must include the management of runaway conditions such as:

- Corrupted program counter
- Unexpected reset
- Critical Data corruption (control registers...)

Prequalification trials

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the Oscillator pins for 1 second.

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

Electromagnetic Interference (EMI)

The electromagnetic field emitted by the device are monitored while a simple application is executed (toggling 2 LEDs through the I/O ports). This emission test is compliant with IEC 61967-2 standard which specifies the test board and the pin loading.

Table 46. EMI characteristics

Symbol	Parameter	Conditions	Monitored	Max vs. [f _{HSE} /f _{HCLK}]	Unit
- Cy201			frequency band	8/48 MHz	
		V 26VT 25°C	0.1 to 30 MHz	-3	
	Peak level	V _{DD} = 3.6 V, T _A = 25 °C, LQFP64 package	30 to 130 MHz	28	dΒμV
S _{EMI}		130 MHz to 1GHz	23		
		01901-2	SAE EMI Level	4	-

7.3.11 Electrical sensitivity characteristics

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts \times (n+1) supply pins). This test conforms to the JESD22-A114/C101 standard.

Table 47. ESD absolute maximum ratings

Symbol	Ratings	Conditions	Class	Maximum value ⁽¹⁾	Unit
V _{ESD(HBM)}	l ~	T _A = +25 °C, conforming to JESD22-A114	2	2000	V
V _{ESD(CDM)}	Electrostatic discharge voltage (charge device model)	T _A = +25 °C, conforming to JESD22-C101	II	500	V

^{1.} Data based on characterization results, not tested in production.

Static latch-up

Two complementary static tests are required on six parts to assess the latch-up performance:

- A supply overvoltage is applied to each power supply pin
- A current injection is applied to each input, output and configurable I/O pin

These tests are compliant with EIA/JESD 78A IC latch-up standard.

Table 48. Electrical sensitivities

Symbol	Parameter	Conditions	Class
LU	Static latch-up class	T _A = +105 °C conforming to JESD78A	II level A

7.3.12 I/O current injection characteristics

As a general rule, current injection to the I/O pins, due to external voltage below V_{SS} or above V_{DD} (for standard, 3 V-capable I/O pins) should be avoided during normal product operation. However, in order to give an indication of the robustness of the microcontroller in cases when abnormal injection accidentally happens, susceptibility tests are performed on a sample basis during device characterization.

Functional susceptibility to I/O current injection

While a simple application is executed on the device, the device is stressed by injecting current into the I/O pins programmed in floating input mode. While current is injected into the I/O pin, one at a time, the device is checked for functional failures.

The failure is indicated by an out of range parameter: ADC error above a certain limit (>5 LSB TUE), out of spec current injection on adjacent pins or other functional failure (for example reset, oscillator frequency deviation).

The test results are given in Table 49

Table 49. I/O current injection susceptibility

		Functional s		
Symbol	Description	Negative injection	Positive injection	Unit
	Injected current on OSC_IN32, OSC_OUT32, PA4, PA5, PC13	TBD	TBD	
	Injected current on all FT pins	TBD	TBD	
I _{INJ}	Injected current on all FTf pins	TBD	TBD	mA
	Injected current on all TTa pins	TBD	TBD	
	Injected current on any other pin	TBD	TBD	

7.3.13 I/O port characteristics

General input/output characteristics

Unless otherwise specified, the parameters given in *Table 50* are derived from tests performed under the conditions summarized in *Table 20*. All I/Os are CMOS and TTL compliant.

Table 50. I/O static characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
	Standard I/O input low level voltage		-0.3		0.3V _{DD} +0.07	
V _{IL}	TTa I/O input low level voltage		-0.3		0.3V _{DD} +0.07	
	FT and FTf ⁽¹⁾ I/O input low level voltage		-0.3		0.475V _{DD} -0.2	V
	Standard I/O input high level voltage		0.445V _{DD} +0.398		V _{DD} +0.3	V
V _{IH}	TTa I/O input high level voltage		0.445V _{DD} +0.398		V _{DD} +0.3	
	FT and FTf ⁽¹⁾ I/O input high level voltage		0.5V _{DD} +0.2		5.5	
	Standard I/O Schmitt trigger voltage hysteresis ⁽²⁾		200			
V _{hys}	TTa I/O Schmitt trigger voltage hysteresis ⁽²⁾		200			mV
	FT and FTf I/O Schmitt trigger voltage hysteresis ⁽²⁾		100			
		$V_{SS} \le V_{IN} \le V_{DD}$ I/O TC, FT and FTf			±0.1	
		$V_{SS} \le V_{IN} \le V_{DD}$ 2 $V \le V_{DD} \le V_{DDA} \le 3.6 \text{ V}$ I/O TTa used in digital mode			±0.1	
	(3)	V _{IN} = 5 V I/O FT and FTf			10	
l _{lkg}	Input leakage current (3)	$\begin{array}{c} V_{IN}{=}~3.6~\text{V},\\ 2~\text{V}{\leq}~\text{V}_{DD}{\leq}~\text{V}_{IN}\\ \text{V}_{DDA}{\;=}~3.6~\text{V}\\ \text{I/O TTa used in digital}\\ \text{mode} \end{array}$			1	μΑ
		$V_{SS} \leq V_{IN} \leq V_{DDA}$ 2 $V \leq V_{DD} \leq V_{DDA} \leq 3.6 \text{ V}$ I/O TTa used in analog mode			±0.2	

Table 50.	I/O static character	istics (continued)
iabie Ju.	I/O Static Cital acte	istics (continucu)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
R _{PU}	Weak pull-up equivalent resistor ⁽⁴⁾	$V_{IN} = V_{SS}$	30	40	50	kΩ
R _{PD}	Weak pull-down equivalent resistor ⁽⁴⁾	$V_{IN} = V_{DD}$	30	40	50	kΩ
C _{IO}	I/O pin capacitance			5		pF

- 1. To sustain a voltage higher than $V_{DD}+0.3$ the internal pull-up/pull-down resistors must be disabled.
- 2. Hysteresis voltage between Schmitt trigger switching levels. Data based on characterization, not tested in production.
- 3. Leakage could be higher than max. if negative current is injected on adjacent pins.
- 4. Pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS/NMOS. This MOS/NMOS contribution to the series resistance is minimum (~10% order).

All I/Os are CMOS and TTL compliant (no software configuration required). Their characteristics cover more than the strict CMOS-technology or TTL parameters. The coverage of these requirements is shown in *Figure 15* and *Figure 16* for standard I/Os, and in *Figure 17* and *Figure 18* for 5 V tolerant I/Os.

Figure 15. TC and TTa I/O input characteristics - CMOS port

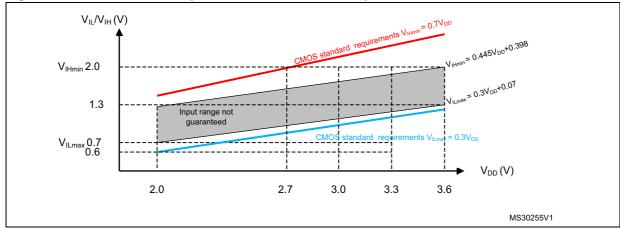


Figure 16. TC and TTa I/O input characteristics - TTL port

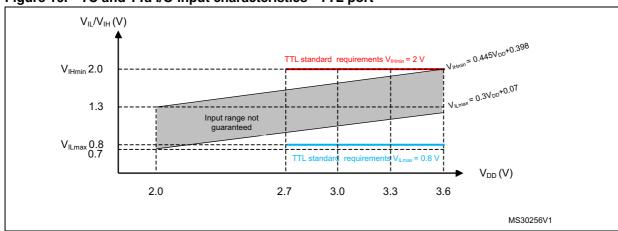


Figure 17. Five volt tolerant (FT and FTf) I/O input characteristics - CMOS port

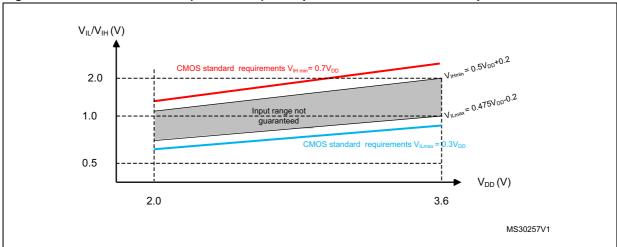
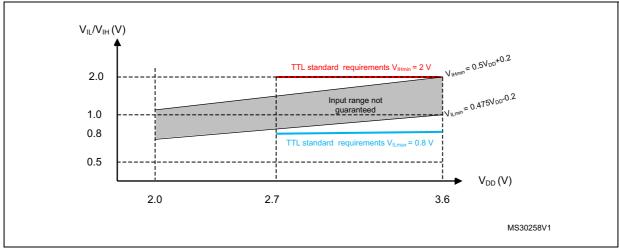


Figure 18. Five volt tolerant (FT and FTf) I/O input characteristics - TTL port



Output driving current

The GPIOs (general purpose input/outputs) can sink or source up to \pm -8 mA, and sink or source up to \pm -20 mA (with a relaxed V_{OL}/V_{OH}).

In the user application, the number of I/O pins which can drive current must be limited to respect the absolute maximum rating specified in *Section 7.2*:

- The sum of the currents sourced by all the I/Os on V_{DD}, plus the maximum Run consumption of the MCU sourced on V_{DD}, cannot exceed the absolute maximum rating I_{VDD} (see *Table 18*).
- The sum of the currents sunk by all the I/Os on V_{SS} plus the maximum Run consumption of the MCU sunk on V_{SS} cannot exceed the absolute maximum rating I_{VSS} (see *Table 18*).

Output voltage levels

Unless otherwise specified, the parameters given in *Table 51* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 20*. All I/Os are CMOS and TTL compliant (FT, TTa or TC unless otherwise specified).

Table 51. Output voltage characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
V _{OL} ⁽¹⁾	Output low level voltage for an I/O pin when 8 pins are sunk at same time	CMOS port ⁽²⁾		0.4	· v
V _{OH} ⁽³⁾	Output high level voltage for an I/O pin when 8 pins are sourced at same time	2.7 V < V _{DD} < 3.6 V	V _{DD} -0.4		
V _{OL} (1)	Output low level voltage for an I/O pin when 8 pins are sunk at same time	TTL port ⁽²⁾		0.4	
V _{OH} ⁽³⁾	Output high level voltage for an I/O pin when 8 pins are sourced at same time	2.7 V < V _{DD} < 3.6 V	2.4		
V _{OL} ⁽¹⁾⁽⁴⁾	Output low level voltage for an I/O pin when 8 pins are sunk at same time	I _{IO} = +20 mA		1.3	- v
V _{OH} ⁽³⁾⁽⁴⁾	Output high level voltage for an I/O pin when 8 pins are sourced at same time	2.7 V < V _{DD} < 3.6 V	V _{DD} -1.3		
V _{OL} ⁽¹⁾⁽⁴⁾	Output low level voltage for an I/O pin when 8 pins are sunk at same time	I _{IO} = +6 mA		0.4	
V _{OH} ⁽³⁾⁽⁴⁾	Output high level voltage for an I/O pin when 8 pins are sourced at same time	2 V < V _{DD} < 2.7 V	V _{DD} -0.4		
V _{OLFM+}	Output low level voltage for an FTf I/O pin in FM+ mode	I _{IO} = +20 mA 2 V < V _{DD} < 3.6 V		0.4	V

^{1.} The $I_{|O}$ current sunk by the device must always respect the absolute maximum rating specified in *Table 18* and the sum of $I_{|O}$ (I/O ports and control pins) must not exceed I_{VSS} .

^{2.} TTL and CMOS outputs are compatible with JEDEC standards JESD36 and JESD52.

^{3.} The $I_{\rm IO}$ current sourced by the device must always respect the absolute maximum rating specified in *Table 18* and the sum of $I_{\rm IO}$ (I/O ports and control pins) must not exceed $I_{\rm VDD}$.

^{4.} Data based on characterization results, not tested in production.

Input/output AC characteristics

The definition and values of input/output AC characteristics are given in *Figure 19* and *Table 52*, respectively.

Unless otherwise specified, the parameters given are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 20*.

Table 52. I/O AC characteristics⁽¹⁾

OSPEEDRy [1:0] value ⁽¹⁾	Symbol	Parameter	Conditions	Min	Max	Unit	
хО	f _{max(IO)out}	Maximum frequency ⁽²⁾	$C_L = 50 \text{ pF}, V_{DD} = 2 \text{ V to } 3.6 \text{ V}$		2	MHz	
	t _{f(IO)out}	Output high to low level fall time	-C _L = 50 pF, V _{DD} = 2 V to 3.6 V		125 ⁽³⁾	- ns	
	t _{r(IO)out}	Output low to high level rise time			125 ⁽³⁾		
	f _{max(IO)out}	Maximum frequency ⁽²⁾	C _L = 50 pF, V _{DD} = 2 V to 3.6 V		10	MHz	
01	t _{f(IO)out}	Output high to low level fall time	C = 50 pE V = 2 V to 2 6 V		25 ⁽³⁾		
	t _{r(IO)out}	Output low to high level rise time	$-C_L = 50 \text{ pF, } V_{DD} = 2 \text{ V to } 3.6 \text{ V}$		25 ⁽³⁾	ns	
	f _{max(IO)out}	Maximum frequency ⁽²⁾	C _L = 30 pF, V _{DD} = 2.7 V to 3.6 V		50	MHz	
			C _L = 50 pF, V _{DD} = 2.7 V to 3.6 V		30	MHz	
			$C_L = 50 \text{ pF}, V_{DD} = 2 \text{ V to } 2.7 \text{ V}$		20	MHz	
	t _{f(IO)out}	Output high to low level fall time	$C_L = 30 \text{ pF}, V_{DD} = 2.7 \text{ V to } 3.6 \text{ V}$		5 ⁽³⁾		
11			C _L = 50 pF, V _{DD} = 2.7 V to 3.6 V		8 ⁽³⁾		
			C _L = 50 pF, V _{DD} = 2 V to 2.7 V		12 ⁽³⁾	٠	
			C _L = 30 pF, V _{DD} = 2.7 V to 3.6 V		5 ⁽³⁾	ns	
	t _{r(IO)out}	Output low to high level rise time	C _L = 50 pF, V _{DD} = 2.7 V to 3.6 V		8 ⁽³⁾		
			C _L = 50 pF, V _{DD} = 2 V to 2.7 V		12 ⁽³⁾		
	f _{max(IO)out}	Maximum frequency ⁽²⁾	TBD		TBD	MHz	
FM+ configuration (4)	t _{f(IO)out}	Output high to low level fall time	TBD		TBD	nc	
	t _{r(IO)out}	Output low to high level rise time	TBD		TBD	ns	
	t _{EXTIpw}	Pulse width of external signals detected by the EXTI controller		10		ns	

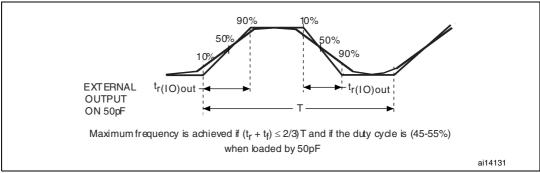
The I/O speed is configured using the OSPEEDRx[1:0] bits. Refer to the RM0091 reference manual for a description of GPIO Port configuration register.

^{2.} The maximum frequency is defined in Figure 19.

^{3.} Guaranteed by design, not tested in production.

The I/O speed configuration is bypassed in FM+ I/O mode. Refer to the STM32F05xxx reference manual RM0091 for a description of FM+ I/O mode configuration.

Figure 19. I/O AC characteristics definition



7.3.14 NRST pin characteristics

The NRST pin input driver uses CMOS technology. It is connected to a permanent pull-up resistor, R_{PLI} (see *Table 50*).

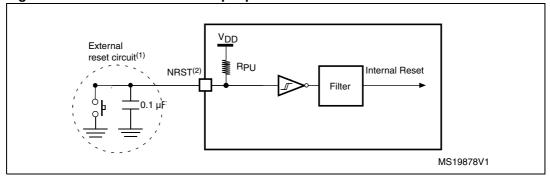
Unless otherwise specified, the parameters given in *Table 53* are derived from tests performed under ambient temperature and VDD supply voltage conditions summarized in *Table 20*.

Table 53. NRST pin characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{IL(NRST)} ⁽¹⁾	NRST Input low level voltage		-0.3		0.8	V
V _{IH(NRST)} ⁽¹⁾	NRST Input high level voltage		2		V _{DD} +0.3	V
V _{hys(NRST)}	NRST Schmitt trigger voltage hysteresis			200		mV
R _{PU}	Weak pull-up equivalent resistor ⁽²⁾	$V_{IN} = V_{SS}$	30	40	50	kΩ
V _{F(NRST)} ⁽¹⁾	NRST Input filtered pulse				100	ns
V _{NF(NRST)} ⁽¹⁾	NRST Input not filtered pulse		300			ns

^{1.} Guaranteed by design, not tested in production.

Figure 20. Recommended NRST pin protection



- The reset network protects the device against parasitic resets.
- The user must ensure that the level on the NRST pin can go below the V_{IL(NRST)} max level specified in Table 53. Otherwise the reset will not be taken into account by the device.

^{2.} The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance must be minimum (\sim 10% order).

7.3.15 Timer characteristics

The parameters given in *Table 54* are guaranteed by design.

Refer to *Section 7.3.13: I/O port characteristics* for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

Table 54. TIMx⁽¹⁾ characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
†(T INA)	Timer resolution time		1		t _{TIMxCLK}
^t res(TIM)	Timer resolution time	f _{TIMxCLK} = 48 MHz	20.8		ns
f _{EXT}	Timer external clock		0	f _{TIMxCLK} /2	MHz
EXI	frequency on CH1 to CH4	f _{TIMxCLK} = 48 MHz	0	24	MHz
Res _{TIM}	Timer resolution	TIMx (except TIM2)		16	bit
TIOSTIM	Timer resolution	TIM2		32	Dit
tCOUNTER	16-bit counter clock period		1	65536	t _{TIMxCLK}
COUNTER	To-bit counter clock period	f _{TIMxCLK} = 48 MHz	0.0208	1365	μs
thank occupit	Maximum possible count			65536 × 65536	t _{TIMxCLK}
^t MAX_COUNT	with 32-bit counter	f _{TIMxCLK} = 48 MHz		89.48	S

TIMx is used as a general term to refer to the TIM1, TIM2, TIM3, TIM6, TIM14, TIM15, TIM16 and TIM17 timers.

Table 55. IWDG min/max timeout period at 40 kHz (LSI) (1)

Prescaler divider	PR[2:0] bits	Min timeout RL[11:0]= 0x000	Max timeout RL[11:0]= 0xFFF	Unit
/4	0	0.1	409.6	
/8	1	0.2	819.2	
/16	2	0.4	1638.4	
/32	3	0.8	3276.8	ms
/64	4	1.6	6553.6	
/128	5	3.2	13107.2	
/256	6 or 7	6.4	26214.4	

These timings are given for a 40 kHz clock but the microcontroller's internal RC frequency can vary from 30 to 60 kHz. Moreover, given an exact RC oscillator frequency, the exact timings still depend on the phasing of the APB interface clock versus the LSI clock so that there is always a full RC period of uncertainty.

Table 56. WWDG min-max timeout value @48 MHz (PCLK)

Prescaler	WDGTB	Min timeout value	Max timeout value	Unit
1	0	0.0853	0.0853 5.4613	
2	1	0.1706	10.9226	mo
4	2	0.3413	21.8453	ms
8	3	0.6826	43.6906	

7.3.16 Communications interfaces

I²C interface characteristics

Unless otherwise specified, the parameters given in *Table 57* are derived from tests performed under ambient temperature, f_{PCLK1} frequency and V_{DD} supply voltage conditions summarized in *Table 20*.

The I^2C interface meets the requirements of the standard I^2C communication protocol with the following restrictions: the I/O pins SDA and SCL are mapped to are not "true" opendrain. When configured as open-drain, the PMOS connected between the I/O pin and V_{DD} is disabled, but is still present.

The I²C characteristics are described in *Table 57*. Refer also to *Section 7.3.13: I/O port characteristics* for more details on the input/output alternate function characteristics (SDA and SCL).

Table 57. I²C characteristics⁽¹⁾

Symbol	Parameter	Standard mod		mode Fast mode		Fast Mode Plus		Unit
Symbol	Parameter	Min	Max	Min	Max	Min	Max	Unit
t _{w(SCLL)}	SCL clock low time	4.7		1.3		0.5		0
t _{w(SCLH)}	SCL clock high time	4.0		0.6		0.26		μs
t _{su(SDA)}	SDA setup time	250		100		50		
t _{h(SDA)}	SDA data hold time	0(3)	3450 ⁽²⁾	0(3)	900 ⁽²⁾	0 ⁽⁴⁾	450 ⁽²⁾	
t _{r(SDA)} t _{r(SCL)}	SDA and SCL rise time		1000		300		120	ns
t _{f(SDA)} t _{f(SCL)}	SDA and SCL fall time		300		300		120	
t _{h(STA)}	Start condition hold time	4.0		0.6		0.26		
t _{su(STA)}	Repeated Start condition setup time	4.7		0.6		0.26		μs
t _{su(STO)}	Stop condition setup time	4.0		0.6		0.26		μS
t _{w(STO:STA)}	Stop to Start condition time (bus free)	4.7		1.3		0.5		μS
C _b	Capacitive load for each bus line		400		400		550	pF

The I2C characteristics are the requirements from I2C bus specification rev03. They are guaranteed by design when I2Cx_TIMING register is correctly programmed (Refer to reference manual). These characteristics are not tested in production

^{2.} The maximum data hold time has only to be met if the interface does not stretch the low period of SCL signal.

^{3.} The device must internally provide a hold time of at least 300ns for the SDA signal in order to bridge the undefined region of the falling edge of SCL.

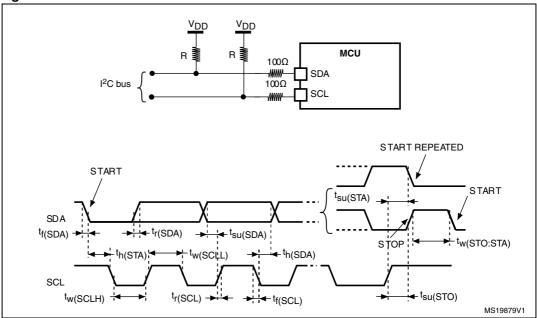
^{4.} The device must internally provide a hold time of at least 120ns for the SDA signal in order to bridge the undefined region of the falling edge of SCL.

Table 58. I2C analog filter characteristics⁽¹⁾

Symbol	Parameter	Min	Max	Unit
t _{SP}	Pulse width of spikes that are suppressed by the analog filter	50	260	ns

^{1.} Guaranteed by design, not tested in production.

Figure 21. I²C bus AC waveforms and measurement circuit



1. Measurement points are done at CMOS levels: $\rm 0.3V_{DD}$ and $\rm 0.7V_{DD}.$

SPI/I²S characteristics

Unless otherwise specified, the parameters given in *Table 59* for SPI or in *Table 60* for I^2S are derived from tests performed under ambient temperature, f_{PCLKx} frequency and V_{DD} supply voltage conditions summarized in *Table 20*.

Refer to *Section 7.3.13: I/O port characteristics* for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO for SPI and WS, CK, SD for I²S).

Table 59. SPI characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
f _{SCK}	SPI clock frequency	Master mode		18	MHz
1/t _{c(SCK)}	SPI Clock frequency	Slave mode		18	IVITZ
t _{r(SCK)}	SPI clock rise and fall time	Capacitive load: C = 15 pF		6	ns
t _{su(NSS)} ⁽¹⁾	NSS setup time	Slave mode	4Tpclk		
t _{h(NSS)} ⁽¹⁾	NSS hold time	Slave mode	2Tpclk + 10		
t _{w(SCKH)} (1) t _{w(SCKL)} (1)	SCK high and low time	Master mode, f _{PCLK} = 36 MHz, presc = 4	Tpclk/2 -2	Tpclk/2 + 1	
t _{su(MI)} (1)	Data input setup time	Master mode	4		
t _{su(MI)} (1) t _{su(SI)} (1)	Data input setup time	Slave mode	5		
t _{h(MI)} (1)	Data input hold time	Master mode	4		
t _{h(SI)} ⁽¹⁾	Data input hold time	Slave mode	5		ns
t _{a(SO)} (1)(2)	Data output access time	Slave mode, f _{PCLK} = 20 MHz	0	3Tpclk	
t _{dis(SO)} (1)(3)	Data output disable time	Slave mode	0	18	
t _{v(SO)} (1)	Data output valid time	Slave mode (after enable edge)		22.5	
t _{v(MO)} ⁽¹⁾	Data output valid time	Master mode (after enable edge)		6	
t _{h(SO)} ⁽¹⁾	Data autaut hald time	Slave mode (after enable edge)	11.5		
t _{h(MO)} ⁽¹⁾	Data output hold time	Master mode (after enable edge)	2		
DuCy(SCK)	SPI slave input clock duty cycle	Slave mode	25	75	%

^{1.} Data based on characterization results, not tested in production.

^{2.} Min time is for the minimum time to drive the output and the max time is for the maximum time to validate the data.

^{3.} Min time is for the minimum time to invalidate the output and the max time is for the maximum time to put the data in Hi-Z

Figure 22. SPI timing diagram - slave mode and CPHA = 0

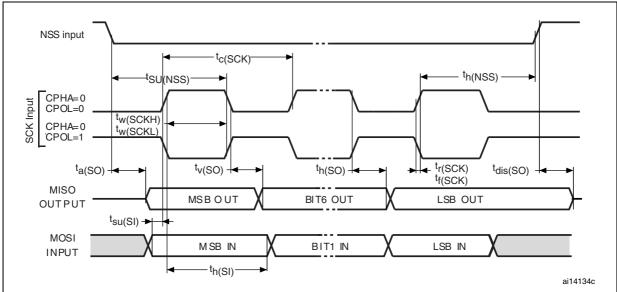
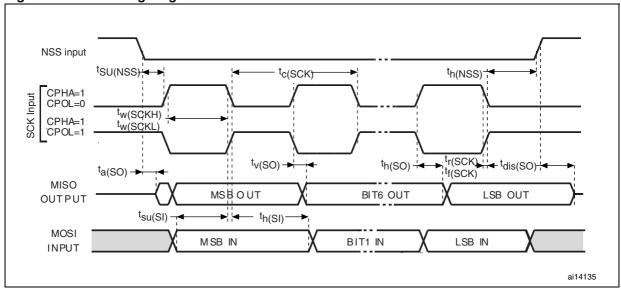
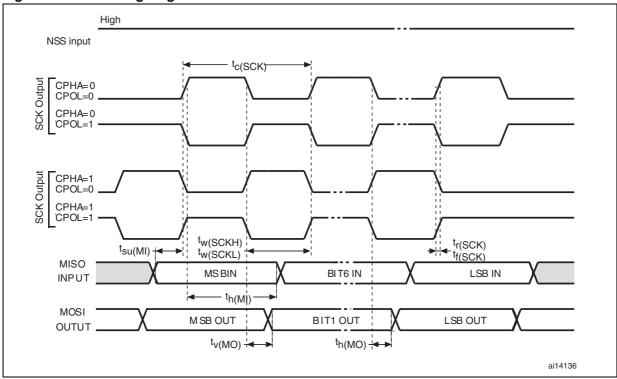


Figure 23. SPI timing diagram - slave mode and CPHA = $1^{(1)}$



1. Measurement points are done at CMOS levels: $0.3V_{DD}$ and $0.7V_{DD}$.

Figure 24. SPI timing diagram - master mode⁽¹⁾



1. Measurement points are done at CMOS levels: $0.3V_{DD}$ and $0.7V_{DD}$.

Table 60. I²S characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
f _{CK}	I ² S clock frequency	Master mode (data: 16 bits, Audio frequency = 48 kHz)	1.597	1.601	MHz
1/t _{c(CK)}		Slave mode	0	6.5	
t _{r(CK)}	I ² S clock rise time	Consolitive load C 15 pF		10	
t _{f(CK)}	I ² S clock fall time	Capacitive load C _L = 15 pF		12	
t _{w(CKH)} (1)	I2S clock high time	Master f _{PCLK} = 16 MHz, audio	306		
t _{w(CKL)} (1)	I2S clock low time	frequency = 48 kHz	312		
t _{v(WS)} (1)	WS valid time	Master mode	2		ns
t _{h(WS)} (1)	WS hold time	Master mode	2		
t _{su(WS)} (1)	WS setup time	Slave mode	7		
t _{h(WS)} (1)	WS hold time	Slave mode	0		
DuCy(SCK)	I2S slave input clock duty cycle	Slave mode	25	75	%
t _{su(SD_MR)} (1)	Data input setup time	Master receiver	6		
t _{su(SD_SR)} (1)	Data input setup time	Slave receiver	2		
t _{h(SD_MR)} ⁽¹⁾⁽²⁾	Data is and bald time	Master receiver	4		
t _{h(SD_SR)} (1)(2)	Data input hold time	Slave receiver	0.5		
t _{v(SD_ST)} (1)(2)	Data output valid time	Slave transmitter (after enable edge)		31 ⁽³⁾	ns
t _{h(SD_ST)} (1)	Data output hold time	Slave transmitter (after enable edge)	13		
t _{v(SD_MT)} (1)(2)	Data output valid time	Master transmitter (after enable edge)		4	
t _{h(SD_MT)} (1)	Data output hold time	Master transmitter (after enable edge)	0		

^{1.} Data based on design simulation and/or characterization results, not tested in production.

^{2.} Depends on f_{PCLK} . For example, if f_{PCLK} =8 MHz, then T_{PCLK} = 1/ f_{PLCLK} =125 ns.

^{3.} REGOFF value was chosen but in REGON target value is 20 ns $\,$

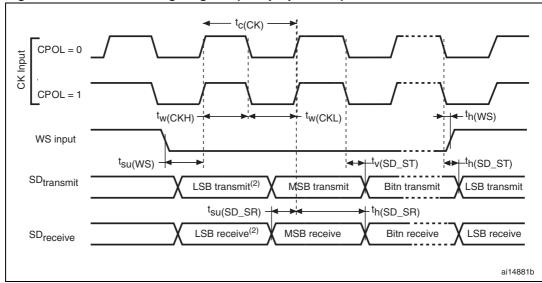


Figure 25. I²S slave timing diagram (Philips protocol)⁽¹⁾

- 1. Measurement points are done at CMOS levels: $0.3 \times V_{DD}$ and $0.7 \times V_{DD}$.
- 2. LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first byte.

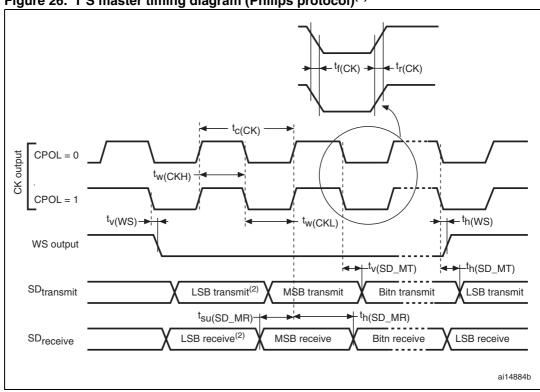


Figure 26. I²S master timing diagram (Philips protocol)⁽¹⁾

- Data based on characterization results, not tested in production.
- LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first byte.

7.3.17 12-bit ADC characteristics

Unless otherwise specified, the parameters given in *Table 61* are preliminary values derived from tests performed under ambient temperature, f_{PCLK2} frequency and V_{DDA} supply voltage conditions summarized in *Table 20*.

Note: It is recommended to perform a calibration after each power-up.

Table 61. ADC characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{DDA}	Analog supply voltage for ADC ON		2.4 3.6		3.6	V
f _{ADC}	ADC clock frequency		0.6		14	MHz
f _S ⁽¹⁾	Sampling rate		0.05		1	MHz
f _{TRIG} ⁽¹⁾	External trigger frequency	f _{ADC} = 14 MHz			823	kHz
'TRIG`	External trigger frequency				17	1/f _{ADC}
V _{AIN}	Conversion voltage range		0		V_{DDA}	V
R _{AIN} ⁽¹⁾	External input impedance	See Equation 1 and Table 62 for details			50	kΩ
R _{ADC} ⁽¹⁾	Sampling switch resistance				1	kΩ
C _{ADC} ⁽¹⁾	Internal sample and hold capacitor				8	pF
t _{CAL} ⁽¹⁾	Calibration time	f _{ADC} = 14 MHz	5.9		μs	
CAL	Cambration time			83		1/f _{ADC}
		$f_{ADC} = f_{PCLK}/2 = 14 \text{ MHz}$	(0.196		μs
		$f_{ADC} = f_{PCLK}/2$		5.5		1/f _{PCLK}
t _{latr} (1)	Trigger conversion latency	$f_{ADC} = f_{PCLK}/4 = 12 \text{ MHz}$		0.219		μs
		$f_{ADC} = f_{PCLK}/4$		10.5		1/f _{PCLK}
		f _{ADC} = f _{HSI14} = 14 MHz	0.188		0.259	μs
Jitter _{ADC}	ADC jitter on trigger conversion	f _{ADC} = f _{HSI14}		1		1/f _{HSI14}
t _S (1)	Compling time	f _{ADC} = 14 MHz	0.107		17.1	μs
'S''	Sampling time		1.5		239.5	1/f _{ADC}
t _{STAB} ⁽¹⁾	Power-up time		0	0	1	μs
	Total conversion time	f _{ADC} = 14 MHz	1		18	μs
t _{CONV} ⁽¹⁾	(including sampling time)		14 to 252 (t _S for successive app			1/f _{ADC}

^{1.} Guaranteed by design, not tested in production.

$$\begin{aligned} & \textbf{Equation 1: R_{AIN} max formula} \\ & R_{AIN} < \frac{T_S}{f_{ADC} \times C_{ADC} \times In(2^{N+2})} - R_{ADC} \end{aligned}$$

The formula above (Equation 1) is used to determine the maximum external impedance allowed for an error below 1/4 of LSB. Here N = 12 (from 12-bit resolution).

 R_{AIN} max for $f_{ADC} = 14$ MHz⁽¹⁾ Table 62.

T _s (cycles)	t _S (μs)	R _{AIN} max (kΩ)
1.5	0.11	0.4
7.5	0.54	5.9
13.5	0.96	11.4
28.5	2.04	25.2
41.5	2.96	37.2
55.5	3.96	50
71.5	5.11	NA
239.5	17.1	NA

^{1.} Guaranteed by design, not tested in production.

ADC accuracy⁽¹⁾⁽²⁾⁽³⁾ Table 63.

Symbol	Parameter	Test conditions	Тур	Max ⁽⁴⁾	Unit
ET	Total unadjusted error		±1.3	±2	
EO	Offset error	f _{PCLK} = 48 MHz,	±1	±1.5	
EG	Gain error	f_{ADC} = 14 MHz, R_{AIN} < 10 kΩ, V_{DDA} = 3 V to 3.6 V	±0.5	±1.5	LSB
ED	Differential linearity error	T _A = 25 °C	±0.7	±1	
EL	Integral linearity error		±0.8	±1.5	
ET	Total unadjusted error		±3.3	±4	
EO	Offset error	f _{PCLK} = 48 MHz,	±1.9	±2.8	
EG	Gain error	f_{ADC} = 14 MHz, R_{AIN} < 10 kΩ, V_{DDA} = 2.7 V to 3.6 V	±2.8	±3	LSB
ED	Differential linearity error	T _A = -40 to 105 °C	±0.7	±1.3	
EL	Integral linearity error		±1.2	±1.7	
ET	Total unadjusted error		±3.3	±4	
EO	Offset error	$f_{PCLK} = 48 \text{ MHz},$ $f_{ADC} = 14 \text{ MHz}, R_{AIN} < 10 \text{ k}\Omega,$ $V_{DDA} = 2.4 \text{ V to } 3.6 \text{ V}$	±1.9	±2.8	
EG	Gain error		±2.8	±3	LSB
ED	Differential linearity error	$T_A = 25 ^{\circ}C$	±0.7	±1.3	
EL	Integral linearity error		±1.2	±1.7	

^{1.} ADC DC accuracy values are measured after internal calibration.

- ADC Accuracy vs. Negative Injection Current: Injecting negative current on any of the standard (non-robust) analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to standard analog pins which may potentially inject negative current.
 Any positive injection current within the limits specified for I_{INJ(PIN)} and ΣI_{INJ(PIN)} in Section 7.3.13 does not affect the ADC accuracy.
- Better performance may be achieved in restricted V_{DDA}, frequency and temperature ranges.
- 4. Data based on characterization results, not tested in production.

Figure 27. ADC accuracy characteristics

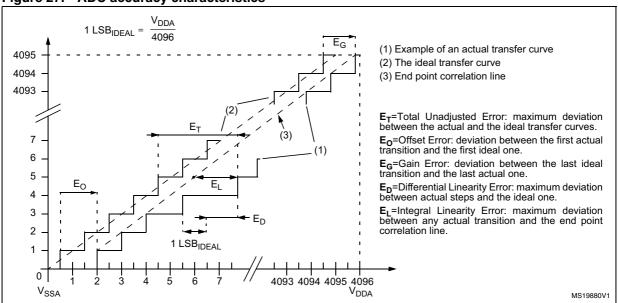
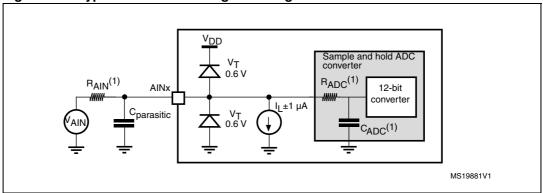


Figure 28. Typical connection diagram using the ADC



- Refer to Table 61 for the values of R_{AIN}, R_{ADC} and C_{ADC}.
- C_{parasitic} represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (roughly 7 pF). A high C_{parasitic} value will downgrade conversion accuracy. To remedy this, f_{ADC} should be reduced.

General PCB design guidelines

Power supply decoupling should be performed as shown in *Figure 9*. The 10 nF capacitor should be ceramic (good quality) and it should be placed as close as possible to the chip.

7.3.18 DAC electrical specifications

Table 64. DAC characteristics

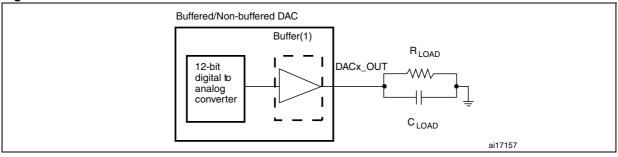
Symbol	Parameter	Min	Тур	Max	Unit	Comments
V _{DDA}	Analog supply voltage for DAC ON	2.4		3.6	٧	
R _{LOAD} ⁽¹⁾	Resistive load with buffer ON	5			kΩ	Load is referred to ground
R _O ⁽¹⁾	Impedance output with buffer OFF			15	kΩ	When the buffer is OFF, the Minimum resistive load between DAC_OUT and V_{SS} to have a 1% accuracy is 1.5 $M\Omega$
C _{LOAD} ⁽¹⁾	Capacitive load			50	pF	Maximum capacitive load at DAC_OUT pin (when the buffer is ON).
DAC_OUT	Lower DAC_OUT voltage with buffer ON	0.2			٧	It gives the maximum output excursion of the DAC. It corresponds to 12-bit input code
DAC_OUT max ⁽¹⁾	Higher DAC_OUT voltage with buffer ON			V _{DDA} - 0.2	V	(0x0E0) to (0xF1C) at V_{DDA} = 3.6 V and (0x155) and (0xEAB) at V_{DDA} = 2.4 V
DAC_OUT min ⁽¹⁾	Lower DAC_OUT voltage with buffer OFF		0.5		mV	It gives the maximum output
DAC_OUT max ⁽¹⁾	Higher DAC_OUT voltage with buffer OFF			V _{DDA} – 1LSB	٧	excursion of the DAC.
1	DAC DC current			380	μΑ	With no load, middle code (0x800) on the input
^I DDA	consumption in quiescent mode (Standby mode)			480	μΑ	With no load, worst code (0xF1C) on the input
DNL ⁽²⁾	Differential non linearity Difference between two			±0.5	LSB	Given for the DAC in 10-bit configuration
	consecutive code-1LSB)			±2	LSB	Given for the DAC in 12-bit configuration
	Integral non linearity (difference between			±1	LSB	Given for the DAC in 10-bit configuration
INL ⁽²⁾	measured value at Code i and the value at Code i on a line drawn between Code 0 and last Code 1023)			±4	I I 🖴 😝	Given for the DAC in 12-bit configuration
	Offset error			±10	mV	Given for the DAC in 12-bit configuration
Offset ⁽²⁾	(difference between measured value at Code (0x800) and the ideal value =			±3		Given for the DAC in 10-bit at $V_{DDA} = 3.6 \text{ V}$
	V _{DDA} /2)			±12	LSB	Given for the DAC in 12-bit at V _{DDA} = 3.6 V
Gain error ⁽²⁾	Gain error			±0.5	%	Given for the DAC in 12bit configuration

Table 64. DAC characteristics (continued)

Symbol	Parameter	Min	Тур	Max	Unit	Comments
t _{SETTLING} (2)	Settling time (full scale: for a 10-bit input code transition between the lowest and the highest input codes when DAC_OUT reaches final value ±1LSB		3	4	μs	$C_{LOAD} \le 50$ pF, $R_{LOAD} \ge 5$ k Ω
Update rate ⁽²⁾	Max frequency for a correct DAC_OUT change when small variation in the input code (from code i to i+1LSB)			1	MS/s	$C_{LOAD} \le 50 \text{ pF, } R_{LOAD} \ge 5 \text{ k}\Omega$
t _{WAKEUP} (2)	Wakeup time from off state (Setting the ENx bit in the DAC Control register)		6.5	10	μs	$C_{LOAD} \leq~50$ pF, $R_{LOAD} \geq 5~k\Omega$ input code between lowest and highest possible ones.
PSRR+ ⁽¹⁾	Power supply rejection ratio (to V _{DDA}) (static DC measurement		-67	-40	dB	No R _{LOAD} , C _{LOAD} = 50 pF

- 1. Guaranteed by design, not tested in production.
- 2. Data based on characterization results, not tested in production.

Figure 29. 12-bit buffered /non-buffered DAC



 The DAC integrates an output buffer that can be used to reduce the output impedance and to drive external loads directly without the use of an external operational amplifier. The buffer can be bypassed by configuring the BOFFx bit in the DAC_CR register.

7.3.19 Comparator characteristics

Table 65. Comparator characteristics

Symbol	Parameter	Condition	ons	Min	Тур	Max ⁽¹⁾	Unit	
V_{DDA}	Analog supply voltage			2		3.6		
V_{IN}	Comparator input voltage range			0		V _{DDA}	V	
V _{BG}	Scaler input voltage				1.2			
V _{SC}	Scaler offset voltage				±5	±10	mV	
t _{S_SC}	Scaler startup time from power down					0.1	ms	
t _{START}	Comparator startup time	Startup time to reach pro specification	pagation delay			60	μs	
		Ultra-low power mode			2	4.5		
	Propagation delay for	Low power mode			0.7	1.5	μs	
	200 mV step with 100	Medium power mode			0.3	0.6	.6	
	mV overdrive	High apped newer made	$V_{DDA} \ge 2.7 \text{ V}$		50	100 240	no	
		High speed power mode	V _{DDA} < 2.7 V		100		115	
t_D		Ultra-low power mode			2	7		
	Propagation delay for	Low power mode			0.7	2.1	μs	
	full range step with 100	Medium power mode			0.3	1.2		
	mV overdrive	High speed power mode	$V_{DDA} \ge 2.7 \text{ V}$		90	180		
		nigh speed power mode	$V_{DDA} < 2.7 \text{ V}$		110	300	ns	
V _{offset}	Comparator offset error	•			±4	±10	mV	
dV _{offset} /dT	Offset error temperature coefficient				18		μV/°C	
		Ultra-low power mode			1.2	1.5		
1	COMP current	Low power mode			3	5		
I _{DD(COMP)}	consumption	Medium power mode			10	15	μA	
		High speed power mode			75	100		

Table 65. Comparator characteristics (continued)

Symbol	Parameter	Conditions			Тур	Max ⁽¹⁾	Unit
		No hysteresis (COMPxHYST[1:0]=00)			0		
V_{hys}		Low hysteresis	High speed power mode	3	8	13	
		(COMPxHYST[1:0]=01)	All other power modes	5	0	10	
	Comparator hysteresis	Medium hysteresis	High speed power mode	7	15	26	mV
		(COMPxHYST[1:0]=10)	All other power modes	9	15	19	
		High hysteresis	High speed power mode	18	31	49	
		(COMPxHYST[1:0]=11)	All other power modes	19	اد	40	

^{1.} Data based on characterization results, not tested in production.

7.3.20 Temperature sensor characteristics

Table 66. TS characteristics

Symbol	Parameter	Min	Тур	Max	Unit
T _L ⁽¹⁾	V _{SENSE} linearity with temperature		±1	±2	°C
Avg_Slope ⁽¹⁾	Average slope	4.0	4.3	4.6	mV/°C
V ₂₅	Voltage at 25 °C	1.34	1.43	1.52	V
t _{START} ⁽¹⁾	Startup time	4		10	μs
T _{S_temp} (1)(2)	ADC sampling time when reading the temperature			17.1	μs

^{1.} Guaranteed by design, not tested in production.

7.3.21 V_{BAT} monitoring characteristics

Table 67. V_{BAT} monitoring characteristics

Symbol	Parameter	Min	Тур	Max	Unit
R	Resistor bridge for V _{BAT}		50		KΩ
Q	Ratio on V _{BAT} measurement		2		
Er ⁽¹⁾	Error on Q	-1		+1	%
T _{S_vbat} ⁽¹⁾⁽²⁾	ADC sampling time when reading the V _{BAT} 1mV accuracy				μs

^{1.} Guaranteed by design, not tested in production.

^{2.} Shortest sampling time can be determined in the application by multiple iterations.

^{2.} Shortest sampling time can be determined in the application by multiple iterations.

8 Package characteristics

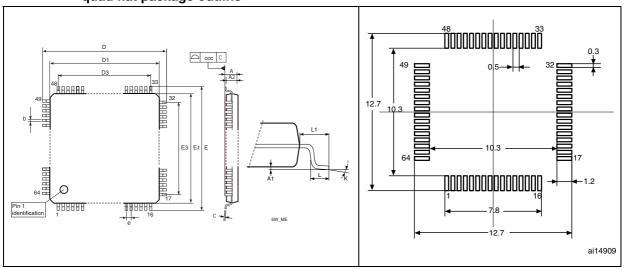
8.1 Package mechanical data

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK® packages, depending on their level of environmental compliance. ECOPACK® specifications, grade definitions and product status are available at: www.st.com. ECOPACK® is an ST trademark.

90/99 Doc ID 022265 Rev 2

Figure 30. LQFP64 – 10 x 10 mm 64 pin low-profile quad flat package outline⁽¹⁾

Figure 31. Recommended footprint (1)(2)



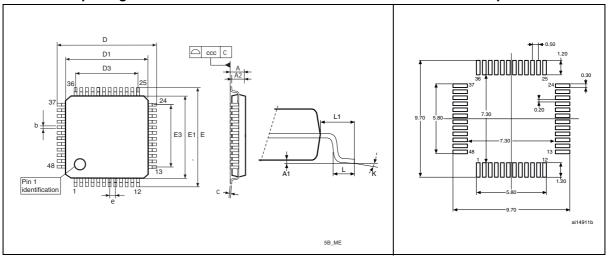
- 1. Drawing is not to scale.
- 2. Dimensions are in millimeters.

Table 68. LQFP64 - 10 x 10 mm 64 pin low-profile quad flat package mechanical data

0		millimeters		inches ⁽¹⁾		
Symbol	Min	Тур	Max	Min	Тур	Max
Α			1.600			0.0630
A1	0.050		0.150	0.0020		0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
С	0.090		0.200	0.0035		0.0079
D	11.800	12.000	12.200	0.4646	0.4724	0.4803
D1	9.800	10.000	10.200	0.3858	0.3937	0.4016
D.		7.500				
E	11.800	12.000	12.200	0.4646	0.4724	0.4803
E1	9.800	10.00	10.200	0.3858	0.3937	0.4016
е		0.500			0.0197	
k	0°	3.5°	7°	0°	3.5°	7°
L	0.450	0.600	0.75	0.0177	0.0236	0.0295
L1		1.000			0.0394	
ccc	0.080 0.0031					
N			Number	r of pins		
			6	4		

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 32. LQFP48 – 7 x 7mm, 48-pin low-profile quad flat Figure 33. Recommended package outline⁽¹⁾ footprint⁽¹⁾⁽²⁾



- 1. Drawing is not to scale.
- 2. Dimensions are in millimeters.

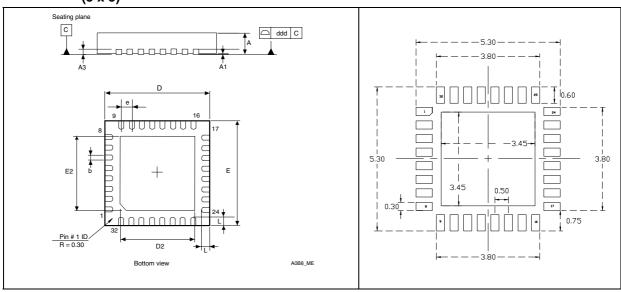
Table 69. LQFP48 - 7 x 7mm, 48-pin low-profile quad flat package mechanical data

Table 05.	EQFF46 - 7 x 7mm, 40-pm low-prome quad nat package mechanical data					
Symbol		millimeters				
Symbol	Min	Тур	Max	Min	Тур	Max
Α			1.600			0.0630
A1	0.050		0.150	0.0020		0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
С	0.090		0.200	0.0035		0.0079
D	8.800	9.000	9.200	0.3465	0.3543	0.3622
D1	6.800	7.000	7.200	0.2677	0.2756	0.2835
D3		5.500			0.2165	
Е	8.800	9.000	9.200	0.3465	0.3543	0.3622
E1	6.800	7.000	7.200	0.2677	0.2756	0.2835
E3		5.500			0.2165	
е		0.500			0.0197	
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1		1.000			0.0394	
k	0°	3.5°	7°	0°	3.5°	7°
ccc		0.080 0.0031				

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 34. UFQFPN32 - 32-lead ultra thin fine pitch quad flat no-lead package outline (5 x 5)⁽¹⁾⁽²⁾⁽³⁾

Figure 35. UFQFPN32 recommended footprint (1)(4)



- 1. Drawing is not to scale.
- 2. All leads/pads should also be soldered to the PCB to improve the lead/pad solder joint life.
- 3. There is an exposed die pad on the underside of the UFQFPN package. This pad is used for the device ground and must be connected. It is referred to as pin 0 in *Table 13: Pin definitions*.
- 4. Dimensions are in millimeters.

Table 70. UFQFPN32 - 32-lead ultra thin fine pitch quad flat no-lead package (5 x 5), package mechanical data

Dim		mm			inches ⁽¹⁾	
Dim.	Min	Тур	Max	Min	Тур	Max
Α	0.5	0.55	0.6	0.0197	0.0217	0.0236
A1	0.00	0.02	0.05	0	0.0008	0.0020
A3		0.152			0.006	
b	0.18	0.23	0.28	0.0071	0.0091	0.0110
D	4.90	5.00	5.10	0.1929	0.1969	0.2008
D2		3.50			0.1378	
E	4.90	5.00	5.10	0.1929	0.1969	0.2008
E2	3.40	3.50	3.60	0.1339	0.1378	0.1417
е		0.500			0.0197	
L	0.30	0.40	0.50	0.0118	0.0157	0.0197
ddd	0.08 0.0031					
	Number of pins					
N	32					

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.

8.2 Thermal characteristics

The maximum chip junction temperature (T_Jmax) must never exceed the values given in *Table 20: General operating conditions on page 42*.

The maximum chip-junction temperature, T_J max, in degrees Celsius, may be calculated using the following equation:

$$T_J \max = T_A \max + (P_D \max x \Theta_{JA})$$

Where:

- T_A max is the maximum ambient temperature in °C,
- Θ_{JA} is the package junction-to-ambient thermal resistance, in °C/W,
- P_D max is the sum of P_{INT} max and $P_{I/O}$ max (P_D max = P_{INT} max + $P_{I/O}$ max),
- P_{INT} max is the product of I_{DD} and V_{DD}, expressed in Watts. This is the maximum chip internal power.

P_{I/O} max represents the maximum power dissipation on output pins where:

$$P_{I/O} \max = \Sigma (V_{OL} \times I_{OL}) + \Sigma ((V_{DD} - V_{OH}) \times I_{OH}),$$

taking into account the actual V_{OL} / I_{OL} and V_{OH} / I_{OH} of the I/Os at low and high level in the application.

Table 71. Package thermal characteristics

Symbol	Parameter	Value	Unit
	Thermal resistance junction-ambient LQFP64 - 10 × 10 mm / 0.5 mm pitch	45	
Θ_{JA}	Thermal resistance junction-ambient LQFP48 - 7 × 7 mm	55	°C/W
	Thermal resistance junction-ambient UFQFPN32 - 5 × 5 mm	38	

8.2.1 Reference document

JESD51-2 Integrated Circuits Thermal Test Method Environment Conditions - Natural Convection (Still Air). Available from www.jedec.org

94/99 Doc ID 022265 Rev 2

8.2.2 Selecting the product temperature range

When ordering the microcontroller, the temperature range is specified in the ordering information scheme shown in *Section 9: Ordering information scheme*.

Each temperature range suffix corresponds to a specific guaranteed ambient temperature at maximum dissipation and, to a specific maximum junction temperature.

As applications do not commonly use the STM32F05xx at maximum dissipation, it is useful to calculate the exact power consumption and junction temperature to determine which temperature range will be best suited to the application.

The following examples show how to calculate the temperature range needed for a given application.

Example 1: High-performance application

Assuming the following application conditions:

Maximum ambient temperature $T_{Amax} = 82$ °C (measured according to JESD51-2), $I_{DDmax} = 50$ mA, $V_{DD} = 3.5$ V, maximum 20 I/Os used at the same time in output at low level with $I_{OL} = 8$ mA, $V_{OL} = 0.4$ V and maximum 8 I/Os used at the same time in output at low level with $I_{OL} = 20$ mA, $V_{OL} = 1.3$ V

 $P_{INTmax} = 50 \text{ mA} \times 3.5 \text{ V} = 175 \text{ mW}$

 $P_{IOmax} = 20 \times 8 \text{ mA} \times 0.4 \text{ V} + 8 \times 20 \text{ mA} \times 1.3 \text{ V} = 272 \text{ mW}$

This gives: P_{INTmax} = 175 mW and P_{IOmax} = 272 mW:

 $P_{Dmax} = 175 + 272 = 447 \text{ mW}$

Using the values obtained in *Table 71* T_{Jmax} is calculated as follows:

For LQFP64, 45 °C/W

 $T_{\text{-lmax}} = 82 \,^{\circ}\text{C} + (45 \,^{\circ}\text{C/W} \times 447 \,^{\circ}\text{mW}) = 82 \,^{\circ}\text{C} + 20.115 \,^{\circ}\text{C} = 102.115 \,^{\circ}\text{C}$

This is within the range of the suffix 6 version parts ($-40 < T_J < 105$ °C) see *Table 20: General operating conditions*.

In this case, parts must be ordered at least with the temperature range suffix 6 (see *Section 9: Ordering information scheme*).

Note:

With this given P_{Dmax} we can find the T_{Amax} allowed for a given device temperature range (order code suffix 6 or 7).

```
Suffix 6: T_{Amax} = T_{Jmax} - (45^{\circ}C/W \times 447 \text{ mW}) = 105-20.115 = 84.885 ^{\circ}C
Suffix 7: T_{Amax} = T_{Jmax} - (45^{\circ}C/W \times 447 \text{ mW}) = 125-20.115 = 104.885 ^{\circ}C
```

Example 2: High-temperature application

Using the same rules, it is possible to address applications that run at high ambient temperatures with a low dissipation, as long as junction temperature T_J remains within the specified range.

Assuming the following application conditions:

Maximum ambient temperature T_{Amax} = 100 °C (measured according to JESD51-2), I_{DDmax} = 20 mA, V_{DD} = 3.5 V, maximum 20 I/Os used at the same time in output at low level with I_{OL} = 8 mA, V_{OL} = 0.4 V

 $P_{INTmax} = 20 \text{ mA} \times 3.5 \text{ V} = 70 \text{ mW}$

 $P_{IOmax} = 20 \times 8 \text{ mA} \times 0.4 \text{ V} = 64 \text{ mW}$

This gives: P_{INTmax} = 70 mW and P_{IOmax} = 64 mW:

 $P_{Dmax} = 70 + 64 = 134 \text{ mW}$

Thus: P_{Dmax} = 134 mW

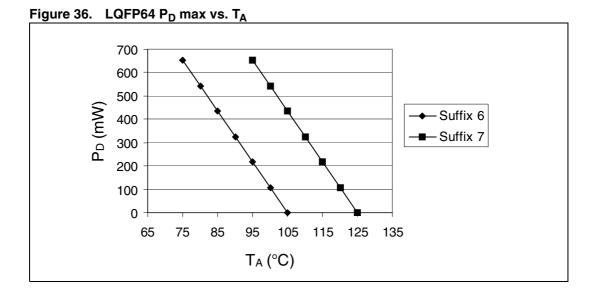
Using the values obtained in $\textit{Table 71}\ T_{\textit{Jmax}}$ is calculated as follows:

$$T_{Jmax} = 100 \, ^{\circ}\text{C} + (45 \, ^{\circ}\text{C/W} \times 134 \, \text{mW}) = 100 \, ^{\circ}\text{C} + 6.03 \, ^{\circ}\text{C} = 106.03 \, ^{\circ}\text{C}$$

This is above the range of the suffix 6 version parts ($-40 < T_J < 105$ °C).

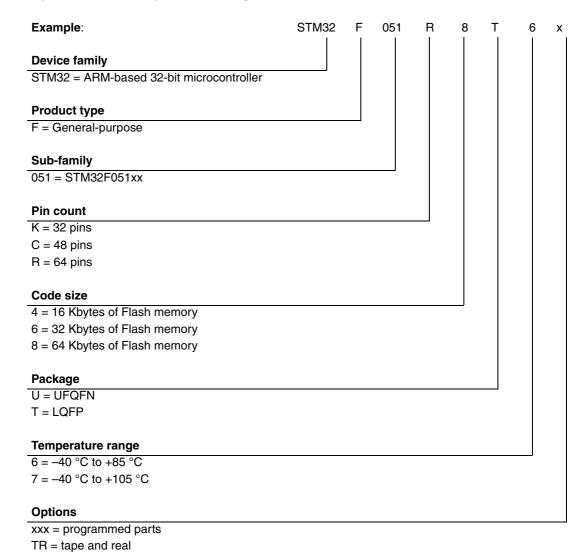
In this case, parts must be ordered at least with the temperature range suffix 7 (see *Section 9: Ordering information scheme*) unless we reduce the power dissipation in order to be able to use suffix 6 parts.

Refer to figure 38 to select the required temperature range (suffix 6 or 7) according to your ambient temperature or power requirements.



9 Ordering information scheme

For a list of available options (memory, package, and so on) or for further information on any aspect of this device, please contact your nearest ST sales office.



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Revision history STM32F051x

10 Revision history

Table 72. Document revision history

Date	Revision	Changes
05-Apr-2012	1	Initial release
25-Apr-2012	2	Updated Table 2: STM32F051xx family device features and peripheral counts for 1 SPI and 1 I2C in 32-pin package Corrected Group 3 pin order in Table 8: Capacitive sensing GPIOs available on STM32F051x devices. Updated current consumption Table 25 to Table 29. Updated Table 39: HSI14 oscillator characteristics

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